

## High-Voltage Ultralow-Iq Low-Dropout Regulator

Check for Samples: [TPS7A6601-Q1](#), [TPS7A6633-Q1](#), [TPS7A6650-Q1](#), [TPS7A6933-Q1](#), [TPS7A6950-Q1](#)

### FEATURES

- Qualified for Automotive Applications
- AEC-Q100 Qualified With the Following Results
  - Device Temperature Grade 1: –40°C to 125°C Ambient Operating Temperature Range
  - Device HBM ESD Classification Level H2
  - Device CDM ESD Classification Level C3B
- 4-V to 40-V Wide Vin Input Voltage Range With up to 45-V Transient
- Output Current 150 mA
- Low Quiescent Current Iq:
  - 2 µA when EN = Low (Shutdown Mode)
  - 20 µA Typical at Light Loads
- Low ESR Ceramic Output Stability Capacitor (2.2 µF–100 µF)
- 250-mV Dropout Voltage at 150 mA (Typical, Vin = 4 V)
- Fixed (3.3-V and 5-V) and Adjustable (1.5-V to 5-V) Output Voltages (Adjustable for TPS7A66xx-Q1 Only)

- Low Input Voltage Tracking
- Integrated Power-On Reset
  - Programmable Reset-Pulse Delay
  - Open-Drain Reset Output
- Integrated Fault Protection
  - Thermal Shutdown
  - Short-Circuit Protection
- Input Voltage Sense Comparator (TPS7A69xx-Q1 Only)
- Packages
  - 8-Pin SOIC-D for TPS7A69xxQ1
  - 8-Pin MSOP-DGN for TPS7A66xx-Q1

### APPLICATIONS

- Qualified for Automotive Applications
- Infotainment Systems With Sleep Mode
- Body Control Modules
- Always-On Battery Applications
  - Gateway Applications
  - Remote Keyless Entry Systems
  - Immobilizers

### DESCRIPTION

The TPS7A66xx and TPS7A69xx are low-dropout linear regulators designed for up to 40-V Vin operations. With only 20-µA quiescent current at no load, they are quite suitable for standby micro control unit systems, especially in automotive applications.

The devices feature integrated short-circuit and overcurrent protection. The devices implement reset delay on power up to indicate the output voltage is stable and in regulation. One can program the delay with an external capacitor. A low-voltage tracking feature allows for a smaller input capacitor and can possibly eliminate the need of using a boost converter during cold-crank conditions.

The devices operate in the –40°C to 125°C temperature range. These features suit the devices well for power supplies in various automotive applications.

**PRODUCT PREVIEW**

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

## TYPICAL APPLICATION SCHEMATIC

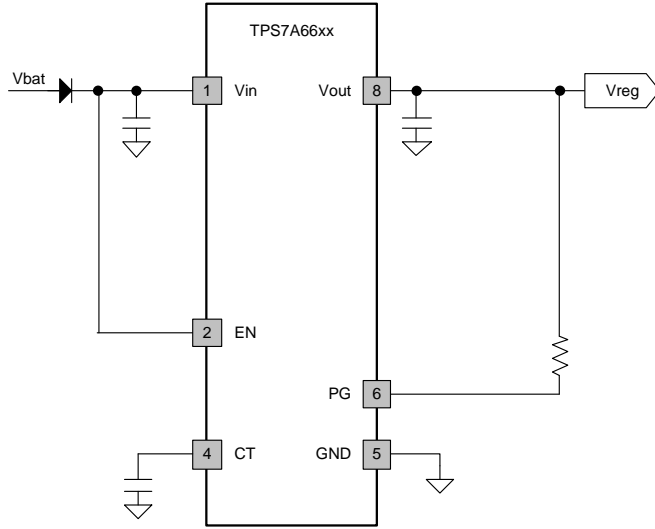


Figure 1. Hardware-Enable Option

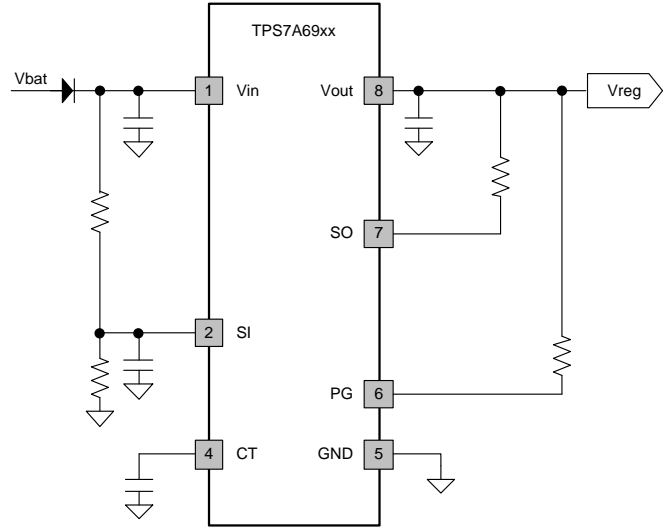


Figure 2. Input-Voltage-Sensing Option



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### ORDERING INFORMATION

PART NUMBER	PACKAGE	VOUT	ORDERABLE PART NUMBER
TPS7A6601-Q1	8 pin MSOP – DGN	Adjustable	TPS7A6601QDGNRQ1
TPS7A6633-Q1	8 pin MSOP – DGN	3.3-V fixed	TPS7A6633QDGNRQ1
TPS7A6650-Q1	8 pin MSOP – DGN	5-V fixed	TPS7A6650QDGNRQ1
TPS7A6933-Q1	8 pin SOIC – D	3.3-V fixed	TPS7A6933QDRQ1
TPS7A6950-Q1	8 pin SOIC – D	5-V fixed	TPS7A6950QDRQ1

### ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted) <sup>(1)</sup>

		MAX	UNITS
V <sub>in</sub> , EN	Unregulated input <sup>(2) (3)(4)</sup>	45	V
V <sub>out</sub>	Regulated output	7	V
SI, CT	See <sup>(2) (3)</sup>	V <sub>in</sub>	V
FB, SO, PG		V <sub>out</sub>	V
ESD	Electrostatic Discharge <sup>(5)</sup>	4	kV
T <sub>A</sub>	Operating ambient temperature range	–40 to 125	°C
T <sub>stg</sub>	Storage temperature range	–65 to 150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values are with respect to GND
- (3) Absolute negative voltage on these pins not to go below –0.3 V
- (4) Absolute maximum voltage, withstand 45 V for 200 ms
- (5) The human-body model is a 107-pF capacitor discharged through a 1.5-kΩ resistor into each pin.

### THERMAL INFORMATION

THERMAL METRIC <sup>(1)</sup>	TPS7A66xx-Q1	TPS7A69xx-Q1	UNITS
	MSOP (8 PINS)	SOIC (8 PINS)	
θ <sub>JA</sub> Junction-to-ambient thermal resistance	63.4	113.2	°C/W
θ <sub>JCtop</sub> Junction-to-case (top) thermal resistance	53.0	59.6	
θ <sub>JB</sub> Junction-to-board thermal resistance <sup>(2)</sup>	37.4	23.4	
ψ <sub>JT</sub> Junction-to-top characterization parameter	3.7	12.8	
ψ <sub>JB</sub> Junction-to-board characterization parameter	37.1	52.9	
θ <sub>JCbot</sub> Junction-to-case (bottom) thermal resistance	13.5	NA	

- (1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).
- (2) The junction-to-board thermal resistance is obtained by simulating in an environment with a ring cold plate fixture to control the PCB temperature, as described in JESD51-8.

## RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNITS
V <sub>in</sub>	Unregulated input	4	40	V
EN, SI, CT		0	40	V
V <sub>out</sub>		1.5	5.5	V
PG, SO, FB	Low voltage (I/O)	0	5.5	V
T <sub>J</sub>	Operating junction temperature range	-40	150	°C

## ELECTRICAL CHARACTERISTICS

V<sub>in</sub> = 14V, 1 mΩ < ESR < 2 Ω, T<sub>J</sub> = -40°C to 150°C (unless otherwise noted)

PARAMETER		CONDITIONS	MIN	TYP	MAX	UNITS
<b>SUPPLY VOLTAGE AND CURRENT(V<sub>in</sub>)</b>						
V <sub>in</sub>	Input voltage	Fixed 5-V output, I <sub>out</sub> = 1 mA	5.5		40	V
		Fixed 3.3-V output, I <sub>out</sub> = 1 mA	4		40	V
I <sub>quiescent</sub>	Quiescent current	V <sub>in</sub> = 5.5 V to 40 V, EN = ON, I <sub>out</sub> = 0.2 mA		20	45	μA
I <sub>Sleep</sub>	Input sleep current	No load current and EN = OFF			4	μA
I <sub>EN</sub>	EN pin current	EN = 40 V			1.0	μA
V <sub>bg</sub>	Band gap	Reference voltage for FB	-2%	1.223	2%	V
V <sub>inUVLO</sub>	Undervoltage detection	Ramp V <sub>in</sub> down until output turns OFF			2.6	V
UVLO <sub>Hys</sub>				1		V
<b>ENABLE INPUT (EN)</b>						
V <sub>IL</sub>	Logic input low level		0		0.4	V
V <sub>IH</sub>	Logic input high level		1.7			V
<b>REGULATED OUTPUT (V<sub>out</sub>)</b>						
V <sub>out</sub>	Regulated output	I <sub>out</sub> = 1 mA, T <sub>J</sub> = 25°C	-1%		1%	
		V <sub>in</sub> = 6 V to 40 V, I <sub>out</sub> = 1 mA to 150 mA <sup>(1)</sup>	-2%		2%	
V <sub>line-reg</sub>	Line regulation	V <sub>in</sub> = 5.5 V to 40 V, ΔV <sub>OUT</sub> , I <sub>out</sub> = 50 mA			10	mV
V <sub>load-reg</sub>	Load regulation	I <sub>out</sub> = 1 mA to 150 mA, ΔV <sub>OUT</sub>			20	mV
V <sub>dropout</sub>	Dropout voltage	V <sub>in</sub> – V <sub>out</sub> , I <sub>out</sub> = 80 mA		133	267	mV
		V <sub>in</sub> – V <sub>out</sub> , I <sub>out</sub> = 150 mA		250	500	
I <sub>out</sub>	Output current	V <sub>out</sub> in regulation	0		150	mA
I <sub>reg-CL</sub>	Output current limit	V <sub>out</sub> short to ground		500		mA
PSRR	Power supply ripple rejection <sup>(2)</sup>	V <sub>in</sub> = 12 V, I <sub>Load</sub> = 10 mA, C <sub>out</sub> = 2.2 μF				
		Freq = 100 Hz		60		dB
		Freq = 100 kHz		40		dB
<b>VOLTAGE SENSING PRE-WARNING</b>						
V <sub>Sith</sub>	Sense low threshold	V <sub>SI</sub> decreasing	1.089	1.123	1.157	V
V <sub>Sith,hys</sub>	Sense threshold hysteresis		20	100	200	mV
V <sub>SOL</sub>	Sense output low voltage	(V <sub>SI</sub> ≤ 1.06 V, V <sub>in</sub> ≥ 4 V, R <sub>SO</sub> = 10 kΩ to V <sub>out</sub> )			0.4	V
I <sub>SOH</sub>	Sense output leakage	(V <sub>SO</sub> = 5 V, V <sub>SI</sub> ≥ 1.5 V)			1	μA
I <sub>SI</sub>	Sense input current		-1	0.1	1	μA

(1) Adjustable version with precision external feedback resistor with tolerance of less than ±1%.

(2) Design information – Not tested, specified by characterization.

**ELECTRICAL CHARACTERISTICS (continued)**

 Vin = 14V, 1 mΩ < ESR < 2 Ω, T<sub>J</sub> = –40°C to 150°C (unless otherwise noted)

PARAMETER		CONDITIONS	MIN	TYP	MAX	UNITS
<b>RESET (PG)</b>						
V <sub>OL</sub>	Reset pulled low	I <sub>OL</sub> = 0.5 mA			0.4	V
I <sub>OH</sub>	Reset pulled Vout through 10-kΩ resistor	Leakage current			1	μA
V <sub>TH-(POR)</sub>	Power-on-reset threshold	Vout power up set tolerance		91.6		% of Vout
V <sub>Thres</sub>	Hysteresis	Vout power down set tolerance		1.6		% of Vout
<b>RESET DELAY (CT)</b>						
I <sub>Chg</sub>	Delay-capacitor charging current	Rdelay = 0 V		1		μA
V <sub>th</sub>	Threshold to release nRST high			1		V
<b>TIMING FOR RESET (PG)</b>						
t <sub>POR</sub>	Power-on-reset delay	Where C = Delay capacitor value Capacitance C = 100 nF <sup>(3)</sup>	50	100	180	ms
t <sub>POR-fixed</sub>		No capacitor on pin	100	290	550	μs
t <sub>Degitch</sub>	Reset deglitch time		20	250		μs
<b>OPERATING TEMPERATURE RANGE</b>						
T <sub>J</sub>	Junction temperature		–40		150	°C
T <sub>shutdown</sub>	Junction shutdown temperature			175		°C
T <sub>Hyst</sub>	Hysteresis of thermal shutdown			20		°C

- (3) This information only will NOT be tested in production and equation will be based as:  $(C \times 1) / 1 \times 10^{-6} = t_{\text{Delay}}$  (delay time).  
Where C = Delay capacitor value. Capacitance C range = 100 pF to 100 nF.

## DEVICE INFORMATION



### Pin Functions

PIN NAME	PIN NO.		TYPE	DESCRIPTION
	SOIC-D	MSOP - DGN		
CT	4	4	O	Reset-pulse delay adjustment. Connecting this pin via a capacitor to GND
EN		2	I	Enable pin. Standby state when enable pin becomes lower than threshold
FB/NC		7	I	Feedback pin when using external resistor divider or NC pin when using internal resistor divider
GND	5	5	G	Ground reference
NC	3	3		Not connected pins
PG	6	6	O	Output ready. This open-drain pin must connect to Vout via an external resistor. The output voltage going below threshold pulls it down.
SI	2		I	Sense input pin to supervise input voltage. Connect via an external voltage divider connected to Vs and GND
SO	7		O	Sense output. This open-drain pin must connect to Vout via an external resistor. The SI voltage becoming lower than the threshold pulls it down.
Vin	1	1	P	Input power-supply voltage
Vout	8	8	P	Output voltage
		—		Thermal pad for MSOP-DGN package

FUNCTIONAL BLOCK DIAGRAMS

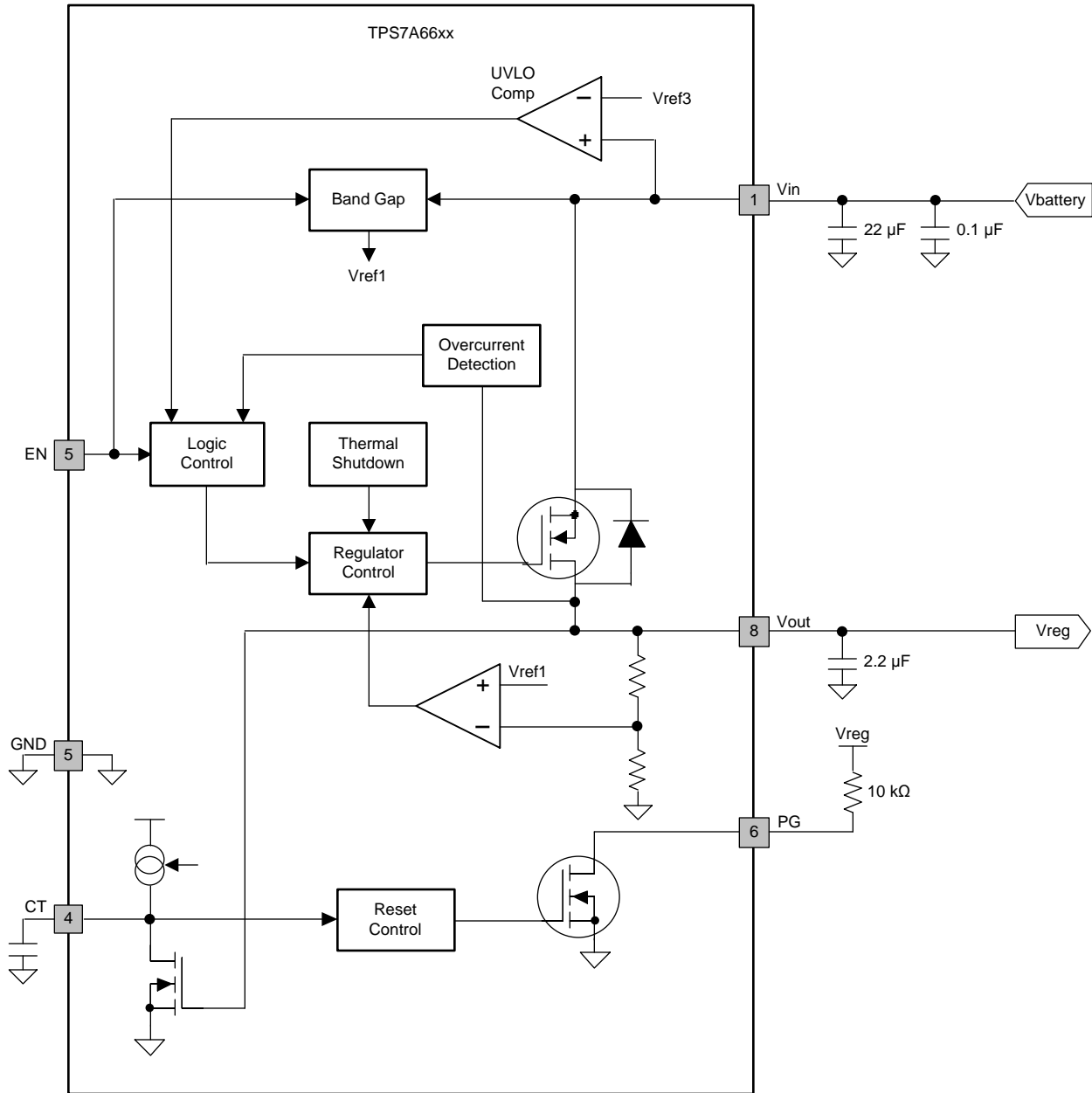


Figure 3. TPS7A66xx Functional Block Diagram

PRODUCT PREVIEW

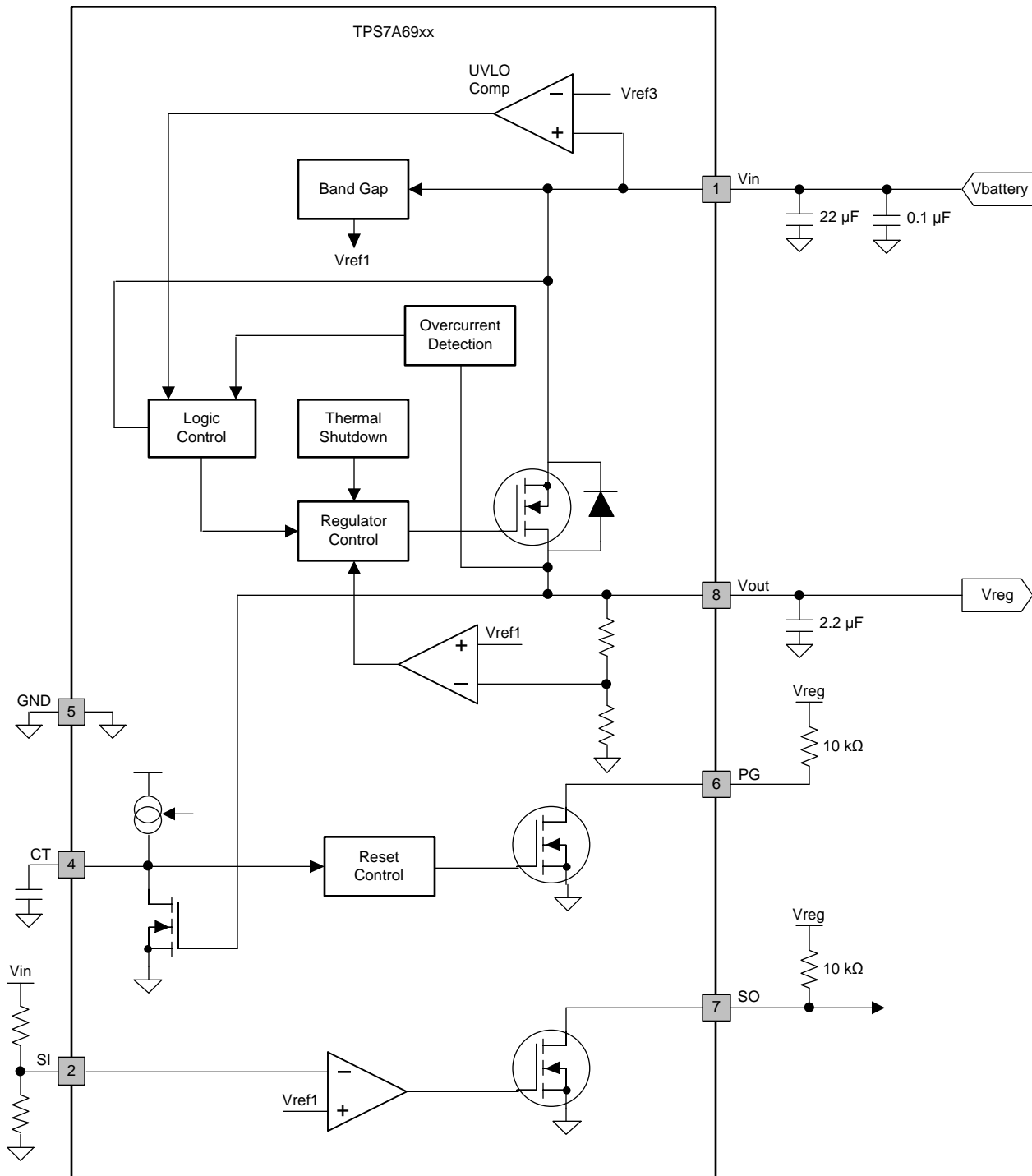


Figure 4. TPS7A69xx Functional Block Diagram

### DETAILED DESCRIPTION

This product is a combination of a low-dropout linear regulator with reset function. The power-on-reset initializes once the output  $V_{out}$  exceeds 91.6% of the target value. The power-on-reset delay is a function of the value set by an external capacitor on the Rdelay pin before releasing the RST terminal high.

#### Enable (EN):

PRODUCT PREVIEW

This is a high-voltage-tolerant terminal; high input activates the device and turns the regulator ON. One can connect this input to the Vin terminal for self-bias applications.

**Regulated Output (Vout):**

This is the regulated output based on the required voltage. The output has current limitation. During initial power up, the regulator has a soft start incorporated to control initial current through the pass element and the output capacitor.

In the event the regulator drops out of regulation, the output tracks the input minus a drop based on the load current. When the input voltage drops below the UVLO threshold, the regulator shuts down until the input voltage recovers above the minimum start-up level.

**Power-On-Reset (PG):**

This is an output with an external pullup resistor to the regulated supply. The output remains low until the regulated Vout has exceeded approximately 90% of the set value and the power-on-reset delay has expired. The on-chip oscillator presets the delay. The regulated output falling below the 90% level asserts this output low after a short de-glitch time of approximately 50 μs (typical).

**Reset Delay Timer (CT):**

An external capacitor on this pin sets the timer delay before the reset pin is asserted high. The constant output current charges an external capacitor until the voltage exceeds a threshold to trip an internal comparator. If this pin is open, the default delay time is 150 μs (typ). After releasing the nRST pin high, the capacitor on this pin discharges, thus allowing the capacitor to charge from approx 0.2 V for the next power-on-reset delay-timer function.

An external capacitor CT defines the reset-pulse delay time, t<sub>CT</sub>, with the charge time of :

$$t_{CT} = \frac{C_{CT} \times 1V}{1\mu A} \tag{1}$$

The power-on-reset initializes once the output Vout exceeds 90% of the programmed value. The power-on-reset delay is a function of the value set by an external capacitor on the CT pin before the releasing of the PG terminal high.

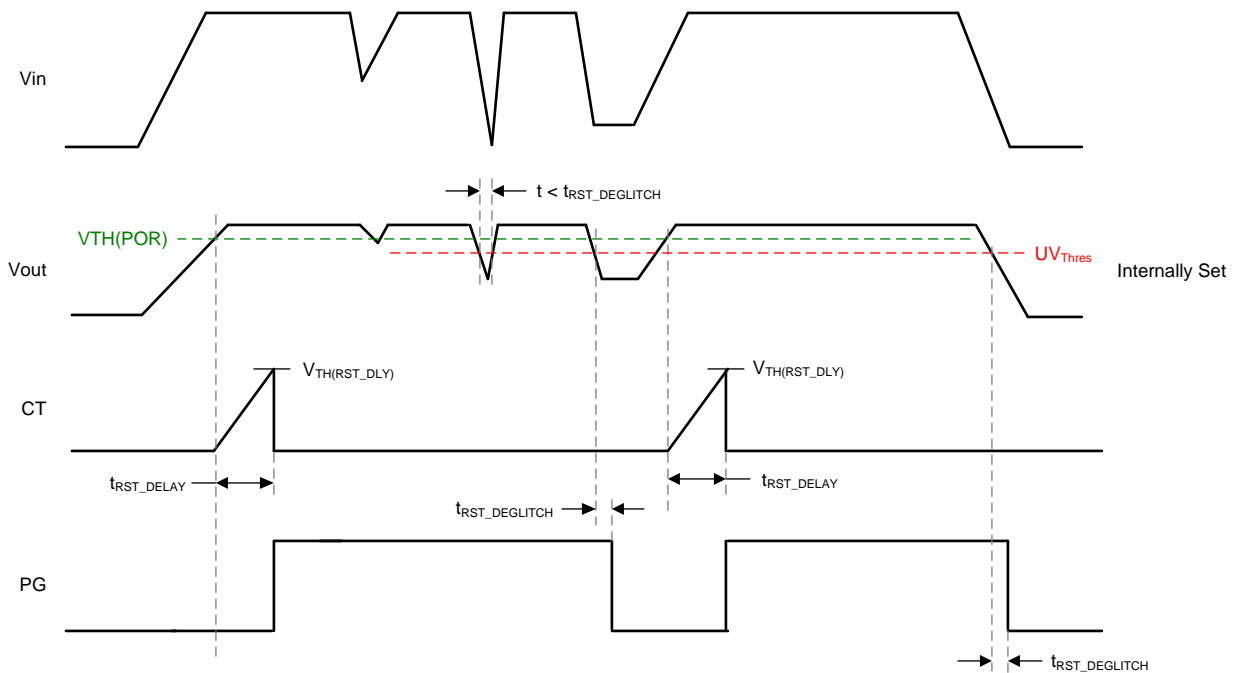


Figure 5. Conditions for Activation of Reset

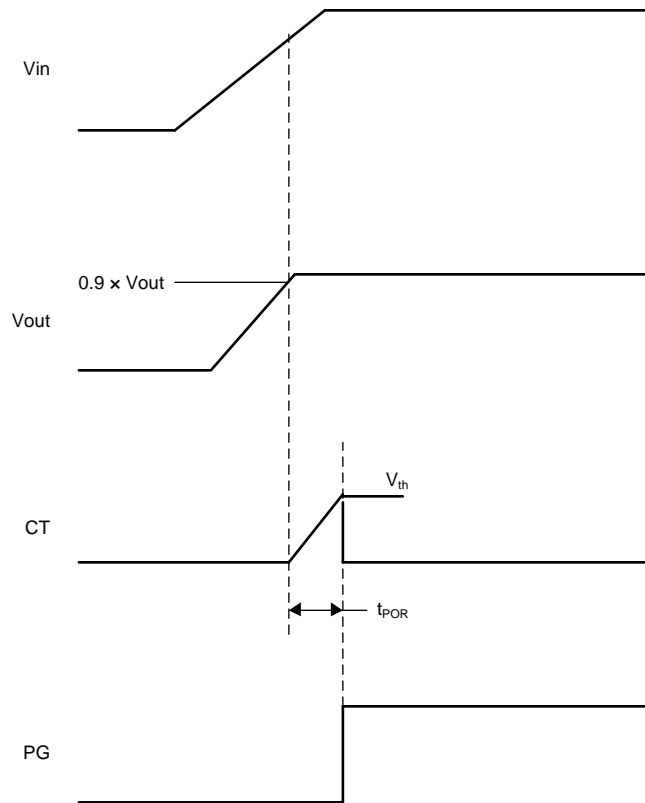


Figure 6. External Programmable Reset Delay

#### Sense Comparator (SI and SO for TPS7A69xx)

The sense comparator compares an input signal with an internal voltage reference of 1.223 V for rising and 1.123 V for falling threshold. The use of an external voltage divider makes this comparator very flexible in the application.

The device can supervise the input voltage either before or after the protection diode and give additional information to the microprocessor, like low-voltage warnings.

The regulator operates in low-power mode when the output load is below 2 mA (typical, 1-mA to 10-mA range). In this mode, the regulator output tolerance is approximately  $V_{out} \pm 1\%$ .

#### Adjustable Output Voltage (FB for TPS7A6601)

One can select an output voltage between 1.5 V and 5.5 V by using the external resistor dividers. Calculate the output voltage using the following equation, where  $V_{FB} = 1.223$  V.

$$V_{out} = V_{FB} \times \left( 1 + \frac{R1}{R2} \right) \quad (2)$$

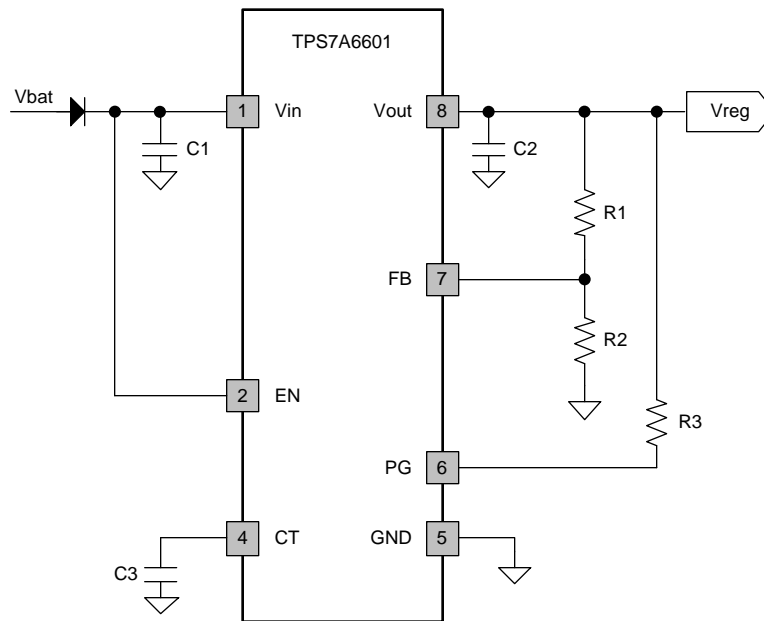


Figure 7. External Feedback Resistor Divider

### Undervoltage Shutdown

There is an internally fixed undervoltage shutdown threshold. Undervoltage shutdown activates when the input voltage on  $V_{in}$  drops below  $V_{in_{UVLO}}$ . This ensures the regulator is not latched into an unknown state during low input supply voltage. If the input voltage has a negative transient which drops below the UVLO threshold and recovers, the regulator shuts down and powers up like a normal power-up sequence once the input voltage is above the required levels.

### Low-Voltage Tracking

At low input voltages the regulator drops out of regulation, the output voltage tracks input minus a voltage based on the load current ( $I_{OUT}$ ) and switch resistance ( $R_{SW}$ ). This allows for a smaller input capacitor and can possibly eliminate the need of using a boost converter during cold-crank conditions.

### Thermal Shutdown

These devices incorporate a thermal shutdown (TSD) circuit as a protection from overheating. For continuous normal operation, the junction temperature should not exceed the TSD trip point. If the junction temperature exceeds the TSD trip point, the output turns off. When the junction temperature falls below the TSD trip point, the output turns on again.

## APPLICATION INFORMATION

Figure 8 and Figure 9 show typical application circuits for the TPS7A66xx and TPS7A69xx, respectively. One may use different values of external components, depending on the end application. An application may require a larger output capacitor may using fast load steps in order to prevent reset from occurring. TI recommends a low-ESR ceramic capacitor with dielectric of type X5R or X7R.

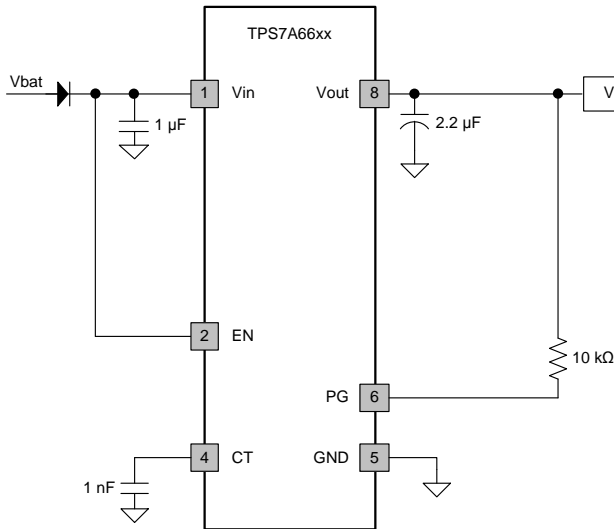


Figure 8. Typical Application Schematic for TPS7A66xx

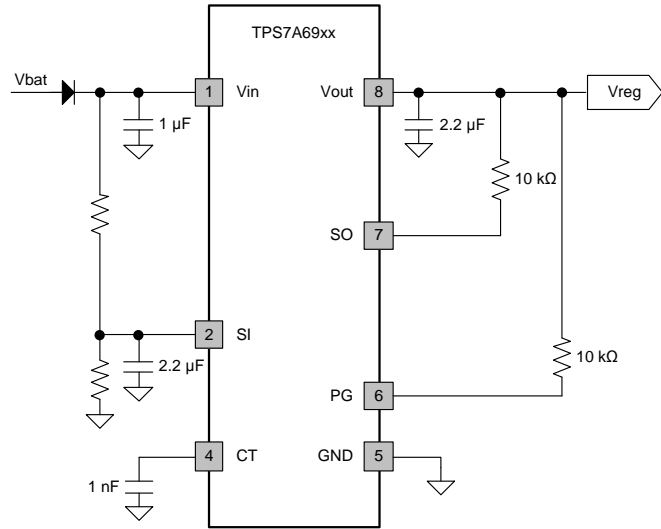


Figure 9. Typical Application Schematic for TPS7A69xx

### Power Dissipation and Thermal Considerations

Calculate power dissipated in the device using Equation 3.

$$P_D = I_{out} \times (V_{in} - V_{out}) + I_{quiescent} \times V_{in} \quad (3)$$

Where:

- $P_D$  = continuous power dissipation
- $I_{out}$  = output current
- $V_{in}$  = input voltage
- $V_{out}$  = output voltage

As  $I_{quiescent} \ll I_{out}$ , therefore ignore the term  $I_{quiescent} \times V_{in}$  in Equation 3.

For a device under operation at a given ambient air temperature ( $T_A$ ), calculate the junction temperature ( $T_J$ ) using Equation 4.

$$T_J = T_A + (\theta_{JA} \times P_D) \quad (4)$$

where:

- $\theta_{JA}$  = junction-to-ambient air thermal impedance

$$\Delta T = T_J - T_A = (\theta_{JA} \times P_D) \quad (5)$$

## LAYOUT INFORMATION

### Package Mounting

Solder pad footprint recommendations for the TPS7A66/69xx-Q1 are available at the end of this product data sheet and at [www.ti.com](http://www.ti.com).

### Board Layout Recommendations to Improve PSRR and Noise Performance

To improve ac performance such as PSRR, output noise, and transient response, TI recommends a board design with separate ground planes for Vin and Vout, with each ground plane connected only at the GND pin of the device. In addition, the ground connection for the output capacitor should connect directly to the GND pin of the device.

Minimize equivalent series inductance (ESL) and ESR in order to maximize performance and ensure stability. Place every capacitor as close as possible to the device and on the same side of the PCB as the regulator itself.

Do not place any of the capacitors on the opposite side of the PCB from where the regulator is installed. TI strongly discourages the use of vias and long traces because they may impact system performance negatively and even cause instability.

If possible, and to ensure the maximum performance specified in this product data sheet, use the same layout pattern used for the TPS7A66/69xx-Q1 evaluation board, available at [www.ti.com](http://www.ti.com).

### Additional Layout Considerations

The high impedance of the FB pin makes the regulator sensitive to parasitic capacitances that may couple undesirable signals from nearby components (especially from logic and digital ICs, such as microcontrollers and microprocessors); these capacitive-coupled signals may produce undesirable output voltage transients. In these cases, TI recommends the use of a fixed-voltage version of the TPS7A66xx-Q1, or isolation of the FB node by flooding the local PCB area with ground-plane copper to minimize any undesirable signal coupling.

### Thermal Protection

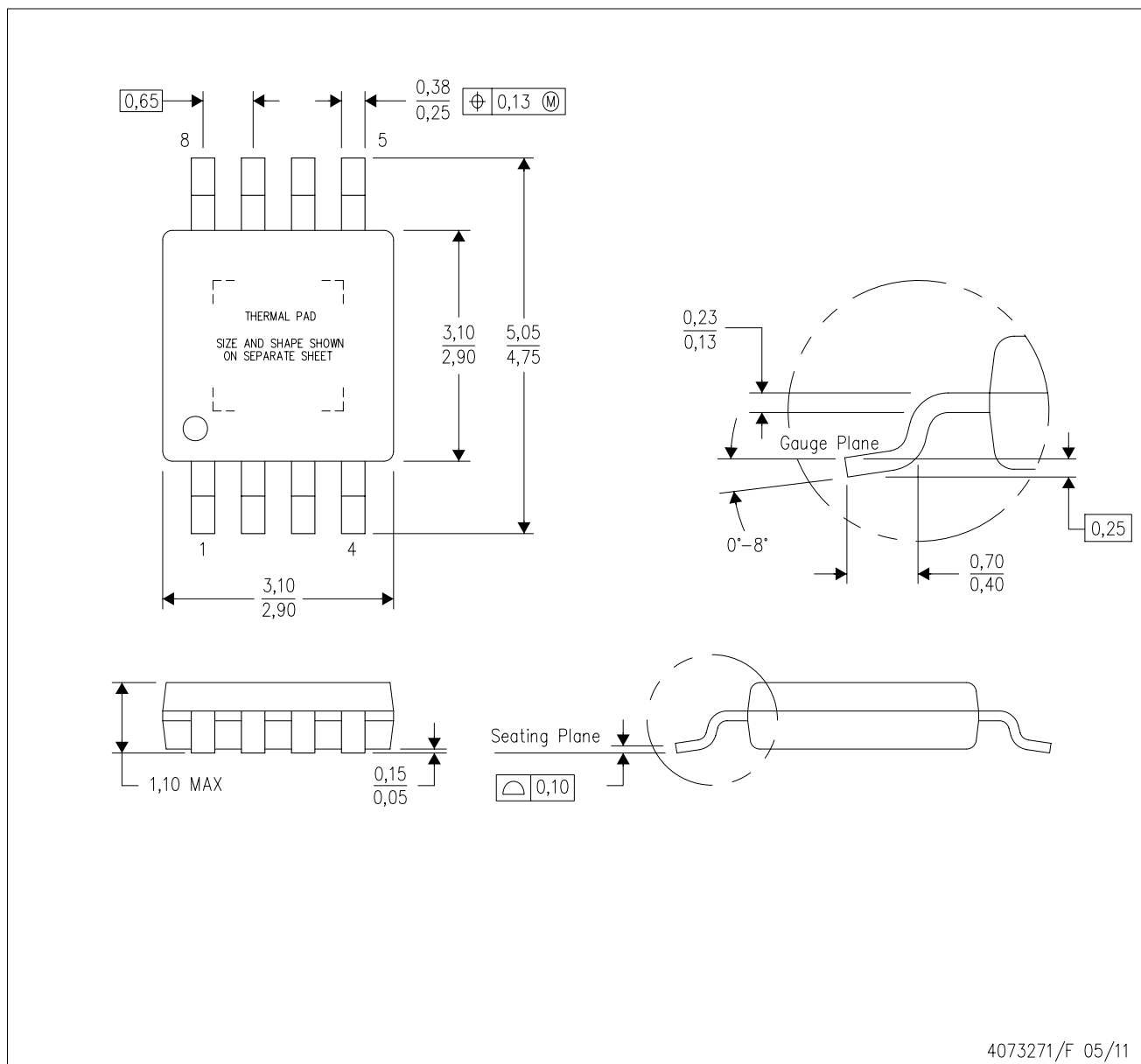
Thermal protection disables the output when the junction temperature rises to approximately 170°C, allowing the device to cool. Cooling of the junction temperature to approximately 150°C enables the output circuitry. Depending on power dissipation, thermal resistance, and ambient temperature, the thermal protection circuit may cycle on and off. This cycling limits the dissipation of the regulator, protecting it from damage as a result of overheating.

Any tendency to activate the thermal protection circuit indicates excessive power dissipation or an inadequate heat-spreading area. For reliable operation, junction temperature should be limited to a maximum of 125°C at the worst-case ambient temperature for a given application. To estimate the margin of safety in a complete design (including the copper heat-spreading area), increase the ambient temperature until the thermal protection is triggered; use worst-case loads and signal conditions. For good reliability, thermal protection should trigger at least 45°C above the maximum expected ambient condition of the particular application. This configuration produces a worst-case junction temperature of 125°C at the highest expected ambient temperature and worst-case load.

The purpose of the design of the internal protection circuitry of the TPS7A66/69xx-Q1 is for protection against overload conditions, not as a replacement for proper heat-sinking. Continuously running the TPS7A66/69xx-Q1 into thermal shutdown degrades device reliability.

DGN (S-PDSO-G8)

PowerPAD™ PLASTIC SMALL OUTLINE

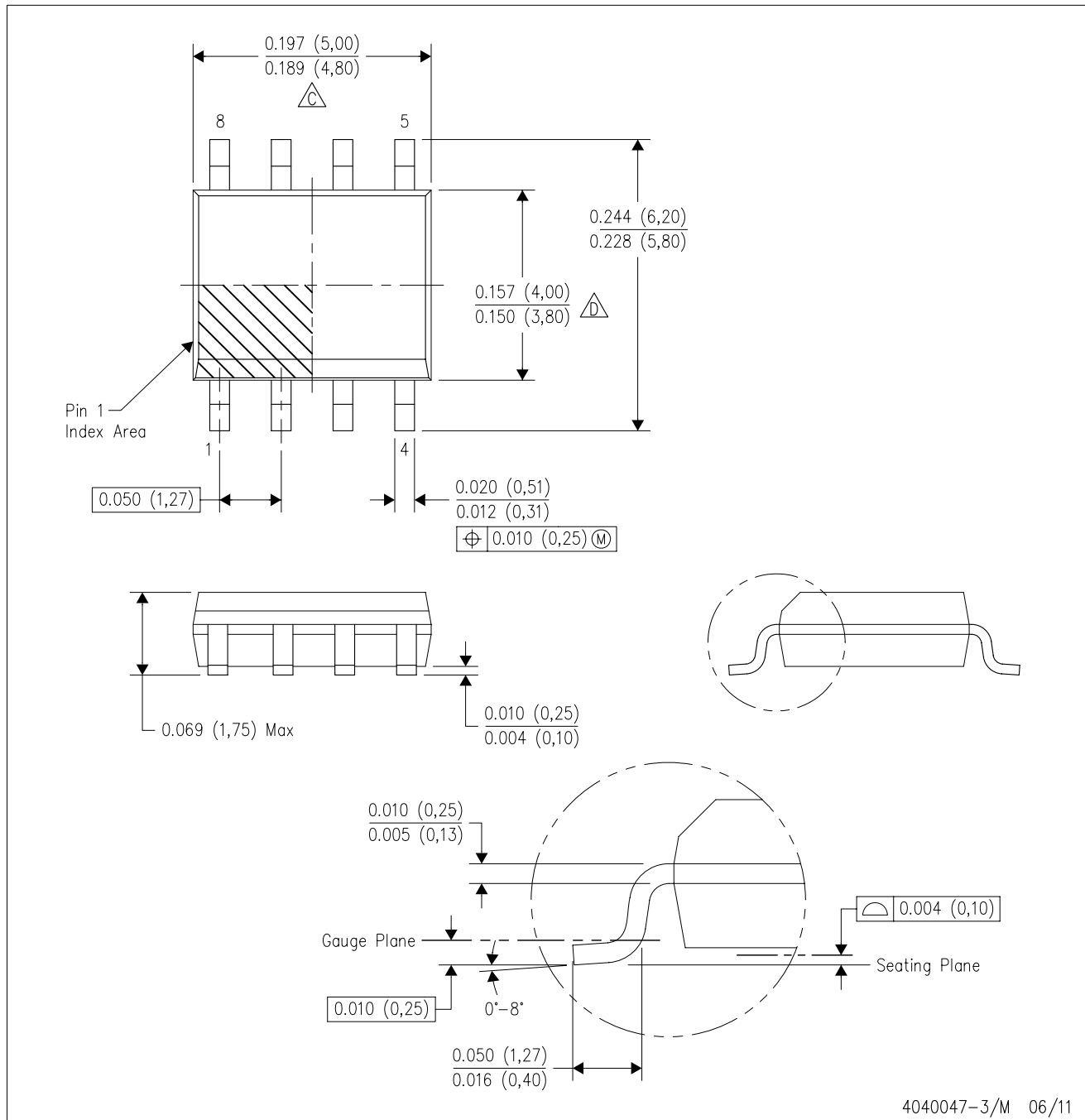


- NOTES:
- All linear dimensions are in millimeters.
  - This drawing is subject to change without notice.
  - Body dimensions do not include mold flash or protrusion.
  - This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at [www.ti.com](http://www.ti.com) <<http://www.ti.com>>.
  - See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
  - Falls within JEDEC MO-187 variation AA-T

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D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  - $\triangle C$  Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
  - $\triangle D$  Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
  - E. Reference JEDEC MS-012 variation AA.

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