

# LM7321/LM7321Q Single/ LM7322/LM7322Q Dual Rail-to-Rail Input/Output $\pm 15V$ , High Output Current and Unlimited Capacitive Load Operational Amplifier

Check for Samples: [LM7321](#), [LM7322](#)

## FEATURES

- ( $V_S = \pm 15V$ ,  $T_A = 25^\circ C$ , Typical Values Unless Specified.)
- Wide Supply Voltage Range 2.5V to 32V
- Output Current +65 mA/–100 mA
- Gain Bandwidth Product 20 MHz
- Slew Rate 18 V/ $\mu s$
- Capacitive Load Tolerance Unlimited
- Input Common Mode Voltage 0.3V Beyond Rails
- Input Voltage Noise 15 nV/ $\sqrt{Hz}$
- Input Current Noise 1.3 pA/ $\sqrt{Hz}$
- Supply Current/Channel 1.1 mA
- Distortion THD+Noise –86 dB
- Temperature Range  $-40^\circ C$  to  $125^\circ C$
- Tested at  $-40^\circ C$ ,  $25^\circ C$  and  $125^\circ C$  at 2.7V,  $\pm 5V$ ,  $\pm 15V$ .
- LM7321Q/LM7322Q are Automotive Grade Products that are AEC-Q100 Grade 1 Qualified.

## APPLICATIONS

- Driving MOSFETs and Power Transistors
- Capacitive Proximity Sensors
- Driving Analog Optocouplers
- High Side Sensing
- Below Ground Current Sensing
- Photodiode Biasing
- Driving Varactor Diodes in PLLs
- Wide Voltage Range Power supplies
- Automotive
- International Power Supplies

## DESCRIPTION

The LM7321/LM7321Q/LM7322/LM7322Q are rail-to-rail input and output amplifiers with wide operating voltages and high output currents. The LM7321/LM7321Q/LM7322/LM7322Q are efficient, achieving 18 V/ $\mu s$  slew rate and 20 MHz unity gain bandwidth while requiring only 1 mA of supply current per op amp. The LM7321/LM7321Q/LM7322/LM7322Q performance is fully specified for operation at 2.7V,  $\pm 5V$  and  $\pm 15V$ .

The LM7321/LM7321Q/LM7322/LM7322Q are designed to drive unlimited capacitive loads without oscillations. All LM7321/LM7321Q and LM7322/LM7322Q parts are tested at  $-40^\circ C$ ,  $125^\circ C$ , and  $25^\circ C$ , with modern automatic test equipment. High performance from  $-40^\circ C$  to  $125^\circ C$ , detailed specifications, and extensive testing makes them suitable for industrial, automotive, and communications applications.

Greater than rail-to-rail input common mode voltage range with 50 dB of common mode rejection across this wide voltage range, allows both high side and low side sensing. Most device parameters are insensitive to power supply voltage, and this makes the parts easier to use where supply voltage may vary, such as automotive electrical systems and battery powered equipment. These amplifiers have true rail-to-rail output and can supply a respectable amount of current (15 mA) with minimal headroom from either rail (300 mV) at low distortion (0.05% THD+Noise). There are several package options for each part. Standard SOIC versions of both parts make upgrading existing designs easy. LM7322/LM7322Q are offered in a space saving 8-Pin VSSOP package. The LM7321/LM7321Q are offered in small SOT-23 package, which makes it easy to place this part close to sensors for better circuit performance.



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## TYPICAL PERFORMANCE CHARACTERISTICS

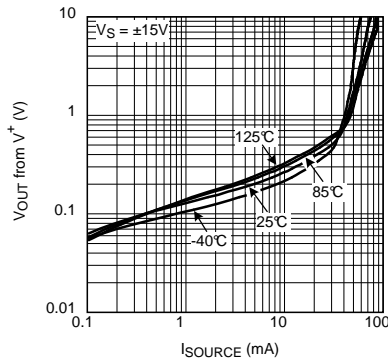


Figure 1. Output Swing vs. Sourcing Current

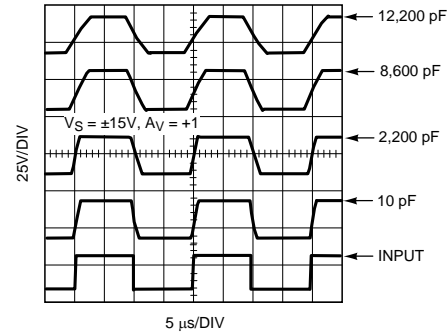


Figure 2. Large Signal Step Response



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### Absolute Maximum Ratings (1)(2)

ESD Tolerance (3)	Human Body Model	2 kV
	Machine Model	200V
	Charge-Device Model	1 kV
VIN Differential		±10V
Output Short Circuit Current		See (4)
Supply Voltage (VS = V+ - V-)		35V
Voltage at Input/Output pins		V+ +0.8V, V- -0.8V
Storage Temperature Range		-65°C to 150°C
Junction Temperature (5)		150°C
Soldering Information:	Infrared or Convection (20 sec.)	235°C
	Wave Soldering (10 sec.)	260°C

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Rating indicate conditions for which the device is intended to be functional, but specific performance is not ensured. For ensured specifications and the test conditions, see the Electrical Characteristics.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (3) Human Body Model, applicable std. MIL-STD-883, Method 3015.7. Machine Model, applicable std. JESD22-A115-A (ESD MM std. of JEDEC) Field-Induced Charge-Device Model, applicable std. JESD22-C101-C (ESD FICDM std. of JEDEC).
- (4) Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C. Short circuit test is a momentary test. Output short circuit duration is infinite for VS ≤ 6V at room temperature and below. For VS > 6V, allowable short circuit duration is 1.5 ms.
- (5) The maximum power dissipation is a function of TJ(MAX), θJA. The maximum allowable power dissipation at any ambient temperature is PD = (TJ(MAX) - TA) / θJA. All numbers apply for packages soldered directly onto a PC board.

### Operating Ratings

Supply Voltage (VS = V+ - V-)		2.5V to 32V
Temperature Range (1)		-40°C to 125°C
Package Thermal Resistance, θJA, (1)	5-Pin SOT-23	325°C/W
	8-Pin VSSOP	235°C/W
	8-Pin SOIC	165°C/W

- (1) The maximum power dissipation is a function of TJ(MAX), θJA. The maximum allowable power dissipation at any ambient temperature is PD = (TJ(MAX) - TA) / θJA. All numbers apply for packages soldered directly onto a PC board.

## 2.7V Electrical Characteristics <sup>(1)</sup>

Unless otherwise specified, all limits ensured for  $T_A = 25^\circ\text{C}$ ,  $V^+ = 2.7\text{V}$ ,  $V^- = 0\text{V}$ ,  $V_{\text{CM}} = 0.5\text{V}$ ,  $V_{\text{OUT}} = 1.35\text{V}$ , and  $R_L > 1\text{M}\Omega$  to  $1.35\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min <sup>(2)</sup>	Typ <sup>(3)</sup>	Max <sup>(2)</sup>	Units
$V_{\text{OS}}$	Input Offset Voltage	$V_{\text{CM}} = 0.5\text{V}$ & $V_{\text{CM}} = 2.2\text{V}$	-5 <b>-6</b>	$\pm 0.7$	+5 <b>+6</b>	mV
TC $V_{\text{OS}}$	Input Offset Voltage Temperature Drift	$V_{\text{CM}} = 0.5\text{V}$ & $V_{\text{CM}} = 2.2\text{V}$ <sup>(4)</sup>		$\pm 2$		$\mu\text{V}/\text{C}$
$I_{\text{B}}$	Input Bias Current	$V_{\text{CM}} = 0.5\text{V}$ <sup>(5)</sup>	-2.0 <b>-2.5</b>	-1.2		$\mu\text{A}$
		$V_{\text{CM}} = 2.2\text{V}$ <sup>(5)</sup>		0.45	1.0 <b>1.5</b>	
$I_{\text{OS}}$	Input Offset Current	$V_{\text{CM}} = 0.5\text{V}$ and $V_{\text{CM}} = 2.2\text{V}$		20	200 <b>300</b>	nA
CMRR	Common Mode Rejection Ratio	$0\text{V} \leq V_{\text{CM}} \leq 1.0\text{V}$	70 <b>60</b>	100		dB
		$0\text{V} \leq V_{\text{CM}} \leq 2.7\text{V}$	55 <b>50</b>	70		
PSRR	Power Supply Rejection Ratio	$2.7\text{V} \leq V_{\text{S}} \leq 30\text{V}$	78 <b>74</b>	104		dB
CMVR	Common Mode Voltage Range	CMRR > 50 dB		-0.3	-0.1 <b>0.0</b>	V
				2.8 <b>2.7</b>	3.0	
$A_{\text{VOL}}$	Open Loop Voltage Gain	$0.5\text{V} \leq V_{\text{O}} \leq 2.2\text{V}$ $R_{\text{L}} = 10\text{ k}\Omega$ to $1.35\text{V}$	65 <b>62</b>	72		dB
		$0.5\text{V} \leq V_{\text{O}} \leq 2.2\text{V}$ $R_{\text{L}} = 2\text{ k}\Omega$ to $1.35\text{V}$	59 <b>55</b>	66		
$V_{\text{OUT}}$	Output Voltage Swing High	$R_{\text{L}} = 10\text{ k}\Omega$ to $1.35\text{V}$ $V_{\text{ID}} = 100\text{ mV}$		50	150 <b>160</b>	mV from either rail
		$R_{\text{L}} = 2\text{ k}\Omega$ to $1.35\text{V}$ $V_{\text{ID}} = 100\text{ mV}$		100	250 <b>280</b>	
	Output Voltage Swing Low	$R_{\text{L}} = 10\text{ k}\Omega$ to $1.35\text{V}$ $V_{\text{ID}} = -100\text{ mV}$		20	120 <b>150</b>	
		$R_{\text{L}} = 2\text{ k}\Omega$ to $1.35\text{V}$ $V_{\text{ID}} = -100\text{ mV}$		40	120 <b>150</b>	
$I_{\text{OUT}}$	Output Current	Sourcing $V_{\text{ID}} = 200\text{ mV}$ , $V_{\text{OUT}} = 0\text{V}$ <sup>(6)</sup>	30 <b>20</b>	48		mA
		Sinking $V_{\text{ID}} = -200\text{ mV}$ , $V_{\text{OUT}} = 2.7\text{V}$ <sup>(6)</sup>	40 <b>30</b>	65		
$I_{\text{S}}$	Supply Current	LM7321		0.95	1.3 <b>1.9</b>	mA
		LM7322		2.0	2.5 <b>3.8</b>	
SR	Slew Rate <sup>(7)</sup>	$A_{\text{V}} = +1$ , $V_{\text{I}} = 2\text{V}$ Step		8.5		$\text{V}/\mu\text{s}$
$f_{\text{u}}$	Unity Gain Frequency	$R_{\text{L}} = 2\text{ k}\Omega$ , $C_{\text{L}} = 20\text{ pF}$		7.5		MHz

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_{\text{J}} = T_{\text{A}}$ . No ensured specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where  $T_{\text{J}} > T_{\text{A}}$ .
- (2) All limits are ensured by testing or statistical analysis.
- (3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.
- (4) Offset voltage temperature drift determined by dividing the change in  $V_{\text{OS}}$  at temperature extremes into the total temperature change.
- (5) Positive current corresponds to current flowing into the device.
- (6) Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of  $150^\circ\text{C}$ . Short circuit test is a momentary test. Output short circuit duration is infinite for  $V_{\text{S}} \leq 6\text{V}$  at room temperature and below. For  $V_{\text{S}} > 6\text{V}$ , allowable short circuit duration is 1.5 ms.
- (7) Slew rate is the slower of the rising and falling slew rates. Connected as a Voltage Follower.

## 2.7V Electrical Characteristics <sup>(1)</sup> (continued)

Unless otherwise specified, all limits ensured for  $T_A = 25^\circ\text{C}$ ,  $V^+ = 2.7\text{V}$ ,  $V^- = 0\text{V}$ ,  $V_{\text{CM}} = 0.5\text{V}$ ,  $V_{\text{OUT}} = 1.35\text{V}$ , and  $R_L > 1\text{M}\Omega$  to  $1.35\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min <sup>(2)</sup>	Typ <sup>(3)</sup>	Max <sup>(2)</sup>	Units
GBW	Gain Bandwidth	$f = 50\text{ kHz}$		16		MHz
$e_n$	Input Referred Voltage Noise Density	$f = 2\text{ kHz}$		11.9		$\text{nV}/\sqrt{\text{Hz}}$
$i_n$	Input Referred Current Noise Density	$f = 2\text{ kHz}$		0.5		$\text{pA}/\sqrt{\text{Hz}}$
THD+N	Total Harmonic Distortion + Noise	$V^+ = 1.9\text{V}$ , $V^- = -0.8\text{V}$ $f = 1\text{ kHz}$ , $R_L = 100\text{ k}\Omega$ , $A_V = +2$ $V_{\text{OUT}} = 210\text{ mV}_{\text{PP}}$		-77		dB
CT Rej.	Crosstalk Rejection	$f = 100\text{ kHz}$ , Driver $R_L = 10\text{ k}\Omega$		60		dB

## $\pm 5\text{V}$ Electrical Characteristics <sup>(1)</sup>

Unless otherwise specified, all limited ensured for  $T_A = 25^\circ\text{C}$ ,  $V^+ = 5\text{V}$ ,  $V^- = -5\text{V}$ ,  $V_{\text{CM}} = 0\text{V}$ ,  $V_{\text{OUT}} = 0\text{V}$ , and  $R_L > 1\text{M}\Omega$  to  $0\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min <sup>(2)</sup>	Typ <sup>(3)</sup>	Max <sup>(2)</sup>	Units
$V_{\text{OS}}$	Input Offset Voltage	$V_{\text{CM}} = -4.5\text{V}$ and $V_{\text{CM}} = 4.5\text{V}$	-5 <b>-6</b>	$\pm 0.7$	+5 <b>+6</b>	mV
TC $V_{\text{OS}}$	Input Offset Voltage Temperature Drift	$V_{\text{CM}} = -4.5\text{V}$ and $V_{\text{CM}} = 4.5\text{V}$ <sup>(4)</sup>		$\pm 2$		$\mu\text{V}/^\circ\text{C}$
$I_B$	Input Bias Current	$V_{\text{CM}} = -4.5\text{V}$ <sup>(5)</sup>	-2.0 <b>-2.5</b>	-1.2		$\mu\text{A}$
		$V_{\text{CM}} = 4.5\text{V}$ <sup>(5)</sup>		0.45	1.0 <b>1.5</b>	
$I_{\text{OS}}$	Input Offset Current	$V_{\text{CM}} = -4.5\text{V}$ and $V_{\text{CM}} = 4.5\text{V}$		20	200 <b>300</b>	nA
CMRR	Common Mode Rejection Ratio	$-5\text{V} \leq V_{\text{CM}} \leq 3\text{V}$	80 <b>70</b>	100		dB
		$-5\text{V} \leq V_{\text{CM}} \leq 5\text{V}$	65 <b>62</b>	80		
PSRR	Power Supply Rejection Ratio	$2.7\text{V} \leq V_S \leq 30\text{V}$ , $V_{\text{CM}} = -4.5\text{V}$	78 <b>74</b>	104		dB
CMVR	Common Mode Voltage Range	CMRR > 50 dB		-5.3	-5.1 <b>-5.0</b>	V
				5.1 <b>5.0</b>	5.3	
$A_{\text{VOL}}$	Open Loop Voltage Gain	$-4\text{V} \leq V_O \leq 4\text{V}$ $R_L = 10\text{ k}\Omega$ to $0\text{V}$	74 <b>70</b>	80		dB
		$-4\text{V} \leq V_O \leq 4\text{V}$ $R_L = 2\text{ k}\Omega$ to $0\text{V}$	68 <b>65</b>	74		

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_J = T_A$ . No ensured specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where  $T_J > T_A$ .
- (2) All limits are ensured by testing or statistical analysis.
- (3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.
- (4) Offset voltage temperature drift determined by dividing the change in  $V_{\text{OS}}$  at temperature extremes into the total temperature change.
- (5) Positive current corresponds to current flowing into the device.

### ±5V Electrical Characteristics <sup>(1)</sup> (continued)

Unless otherwise specified, all limited ensured for  $T_A = 25^\circ\text{C}$ ,  $V^+ = 5\text{V}$ ,  $V^- = -5\text{V}$ ,  $V_{\text{CM}} = 0\text{V}$ ,  $V_{\text{OUT}} = 0\text{V}$ , and  $R_L > 1\text{M}\Omega$  to  $0\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min <sup>(2)</sup>	Typ <sup>(3)</sup>	Max <sup>(2)</sup>	Units
$V_{\text{OUT}}$	Output Voltage Swing High	$R_L = 10\text{ k}\Omega$ to $0\text{V}$ $V_{\text{ID}} = 100\text{ mV}$		100	250 <b>280</b>	mV from either rail
		$R_L = 2\text{ k}\Omega$ to $0\text{V}$ $V_{\text{ID}} = 100\text{ mV}$		160	350 <b>450</b>	
	Output Voltage Swing Low	$R_L = 10\text{ k}\Omega$ to $0\text{V}$ $V_{\text{ID}} = -100\text{ mV}$		35	200 <b>250</b>	
		$R_L = 2\text{ k}\Omega$ to $0\text{V}$ $V_{\text{ID}} = -100\text{ mV}$		80	200 <b>250</b>	
$I_{\text{OUT}}$	Output Current	Sourcing $V_{\text{ID}} = 200\text{ mV}$ , $V_{\text{OUT}} = -5\text{V}$ <sup>(6)</sup>	35 <b>20</b>	70		mA
		Sinking $V_{\text{ID}} = -200\text{ mV}$ , $V_{\text{OUT}} = 5\text{V}$ <sup>(6)</sup>	50 <b>30</b>	85		
$I_S$	Supply Current	$V_{\text{CM}} = -4.5\text{V}$	LM7321		1.0 <b>2</b>	mA
			LM7322		2.3 <b>3.8</b>	
SR	Slew Rate <sup>(7)</sup>	$A_V = +1$ , $V_I = 8\text{V}$ Step		12.3		V/ $\mu\text{s}$
$f_u$	Unity Gain Frequency	$R_L = 2\text{ k}\Omega$ , $C_L = 20\text{ pF}$		9		MHz
GBW	Gain Bandwidth	$f = 50\text{ kHz}$		16		MHz
$e_n$	Input Referred Voltage Noise Density	$f = 2\text{ kHz}$		14.3		nV/ $\sqrt{\text{Hz}}$
$i_n$	Input Referred Current Noise Density	$f = 2\text{ kHz}$		1.35		pA/ $\sqrt{\text{Hz}}$
THD+N	Total Harmonic Distortion + Noise	$f = 1\text{ kHz}$ , $R_L = 100\text{ k}\Omega$ , $A_V = +2$ $V_{\text{OUT}} = 8\text{ V}_{\text{PP}}$		-79		dB
CT Rej.	Crosstalk Rejection	$f = 100\text{ kHz}$ , Driver $R_L = 10\text{ k}\Omega$		60		dB

(6) Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of  $150^\circ\text{C}$ . Short circuit test is a momentary test. Output short circuit duration is infinite for  $V_S \leq 6\text{V}$  at room temperature and below. For  $V_S > 6\text{V}$ , allowable short circuit duration is 1.5 ms.

(7) Slew rate is the slower of the rising and falling slew rates. Connected as a Voltage Follower.

### ±15V Electrical Characteristics <sup>(1)</sup>

Unless otherwise specified, all limited ensured for  $T_A = 25^\circ\text{C}$ ,  $V^+ = 15\text{V}$ ,  $V^- = -15\text{V}$ ,  $V_{\text{CM}} = 0\text{V}$ ,  $V_{\text{OUT}} = 0\text{V}$ , and  $R_L > 1\text{M}\Omega$  to  $15\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min <sup>(2)</sup>	Typ <sup>(3)</sup>	Max <sup>(2)</sup>	Units
$V_{\text{OS}}$	Input Offset Voltage	$V_{\text{CM}} = -14.5\text{V}$ and $V_{\text{CM}} = 14.5\text{V}$	-6 <b>-8</b>	$\pm 0.7$	+6 <b>+8</b>	mV
TC $V_{\text{OS}}$	Input Offset Voltage Temperature Drift	$V_{\text{CM}} = -14.5\text{V}$ and $V_{\text{CM}} = 14.5\text{V}$ <sup>(4)</sup>		$\pm 2$		$\mu\text{V}/^\circ\text{C}$
$I_B$	Input Bias Current	$V_{\text{CM}} = -14.5\text{V}$ <sup>(5)</sup>	-2 <b>-2.5</b>	-1.1		$\mu\text{A}$
		$V_{\text{CM}} = 14.5\text{V}$ <sup>(5)</sup>		0.45	1.0 <b>1.5</b>	
$I_{\text{OS}}$	Input Offset Current	$V_{\text{CM}} = -14.5\text{V}$ and $V_{\text{CM}} = 14.5\text{V}$		30	300 <b>500</b>	nA

(1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_J = T_A$ . No ensured specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where  $T_J > T_A$ .

(2) All limits are ensured by testing or statistical analysis.

(3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.

(4) Offset voltage temperature drift determined by dividing the change in  $V_{\text{OS}}$  at temperature extremes into the total temperature change.

(5) Positive current corresponds to current flowing into the device.

### ±15V Electrical Characteristics <sup>(1)</sup> (continued)

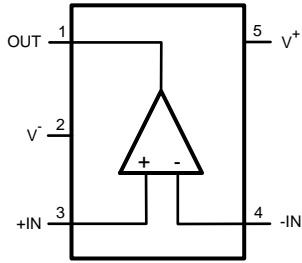
Unless otherwise specified, all limited ensured for  $T_A = 25^\circ\text{C}$ ,  $V^+ = 15\text{V}$ ,  $V^- = -15\text{V}$ ,  $V_{\text{CM}} = 0\text{V}$ ,  $V_{\text{OUT}} = 0\text{V}$ , and  $R_L > 1\text{M}\Omega$  to 15V. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min (2)	Typ (3)	Max (2)	Units
CMRR	Common Mode Rejection Ratio	$-15\text{V} \leq V_{\text{CM}} \leq 12\text{V}$	80 <b>75</b>	100		dB
		$-15\text{V} \leq V_{\text{CM}} \leq 15\text{V}$	72 <b>70</b>	80		
PSRR	Power Supply Rejection Ratio	$2.7\text{V} \leq V_S \leq 30\text{V}$ , $V_{\text{CM}} = -14.5\text{V}$	78 <b>74</b>	100		dB
CMVR	Common Mode Voltage Range	CMRR > 50 dB		-15.3	-15.1 <b>-15</b>	V
				15.1 <b>15</b>	15.3	
A <sub>VOL</sub>	Open Loop Voltage Gain	$-13\text{V} \leq V_O \leq 13\text{V}$ $R_L = 10\text{ k}\Omega$ to 0V	75 <b>70</b>	85		dB
		$-13\text{V} \leq V_O \leq 13\text{V}$ $R_L = 2\text{ k}\Omega$ to 0V	70 <b>65</b>	78		
V <sub>OUT</sub>	Output Voltage Swing High	$R_L = 10\text{ k}\Omega$ to 0V $V_{\text{ID}} = 100\text{ mV}$		150	300 <b>350</b>	mV from either rail
		$R_L = 2\text{ k}\Omega$ to 0V $V_{\text{ID}} = 100\text{ mV}$		250	550 <b>650</b>	
	Output Voltage Swing Low	$R_L = 10\text{ k}\Omega$ to 0V $V_{\text{ID}} = -100\text{ mV}$		60	200 <b>250</b>	
		$R_L = 2\text{ k}\Omega$ to 0V $V_{\text{ID}} = -100\text{ mV}$		130	300 <b>400</b>	
I <sub>OUT</sub>	Output Current	Sourcing $V_{\text{ID}} = 200\text{ mV}$ , $V_{\text{OUT}} = -15\text{V}$ <sup>(6)</sup>	40	65		mA
		Sinking $V_{\text{ID}} = -200\text{ mV}$ , $V_{\text{OUT}} = 15\text{V}$ <sup>(6)</sup>	60	100		
I <sub>S</sub>	Supply Current	$V_{\text{CM}} = -14.5\text{V}$	LM7321	1.1	1.7 <b>2.4</b>	mA
			LM7322	2.5	4 <b>5.6</b>	
SR	Slew Rate <sup>(7)</sup>	$A_V = +1$ , $V_I = 20\text{V}$ Step		18		V/ $\mu\text{s}$
f <sub>u</sub>	Unity Gain Frequency	$R_L = 2\text{ k}\Omega$ , $C_L = 20\text{ pF}$		11.3		MHz
GBW	Gain Bandwidth	$f = 50\text{ kHz}$		20		MHz
e <sub>n</sub>	Input Referred Voltage Noise Density	$f = 2\text{ kHz}$		15		nV/ $\sqrt{\text{Hz}}$
i <sub>n</sub>	Input Referred Current Noise Density	$f = 2\text{ kHz}$		1.3		pA/ $\sqrt{\text{Hz}}$
THD+N	Total Harmonic Distortion +Noise	$f = 1\text{ kHz}$ , $R_L = 100\text{ k}\Omega$ , $A_V = +2$ , $V_{\text{OUT}} = 23\text{ V}_{\text{PP}}$		-86		dB
CT Rej.	Crosstalk Rejection	$f = 100\text{ kHz}$ , Driver $R_L = 10\text{ k}\Omega$		60		dB

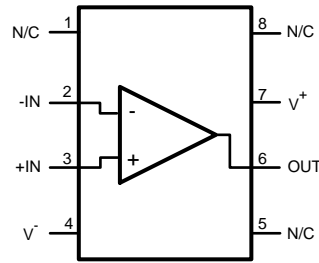
(6) Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C. Short circuit test is a momentary test. Output short circuit duration is infinite for  $V_S \leq 6\text{V}$  at room temperature and below. For  $V_S > 6\text{V}$ , allowable short circuit duration is 1.5 ms.

(7) Slew rate is the slower of the rising and falling slew rates. Connected as a Voltage Follower.

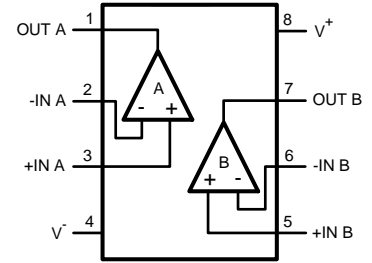
**CONNECTION DIAGRAMS**



**Figure 3. 5-Pin SOT-23  
Top View**



**Figure 4. 8-Pin SOIC  
Top View**



**Figure 5. 8-Pin VSSOP/SOIC  
Top View**

### Typical Performance Characteristics

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

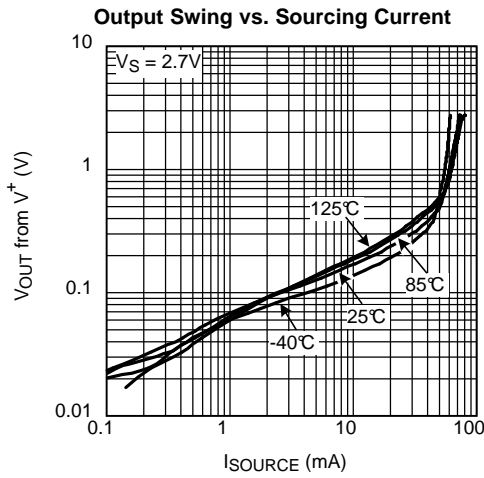


Figure 6.

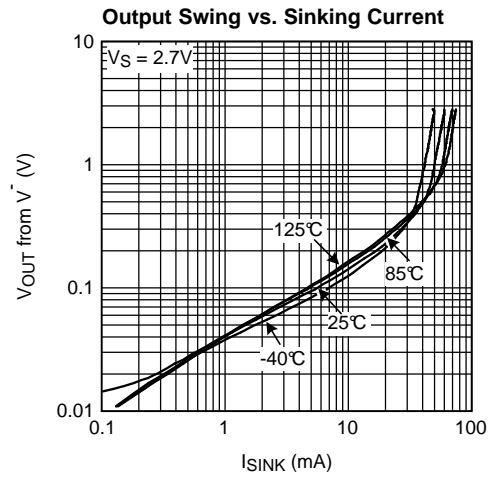


Figure 7.

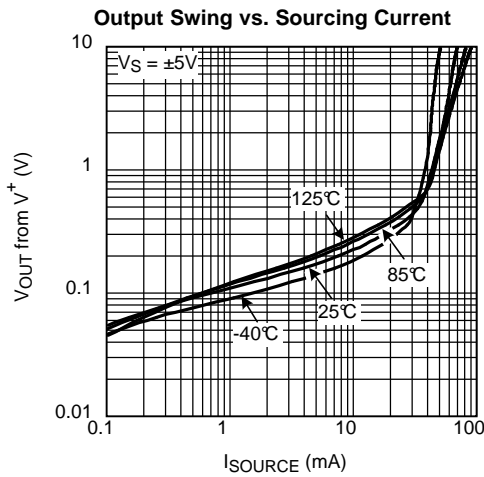


Figure 8.

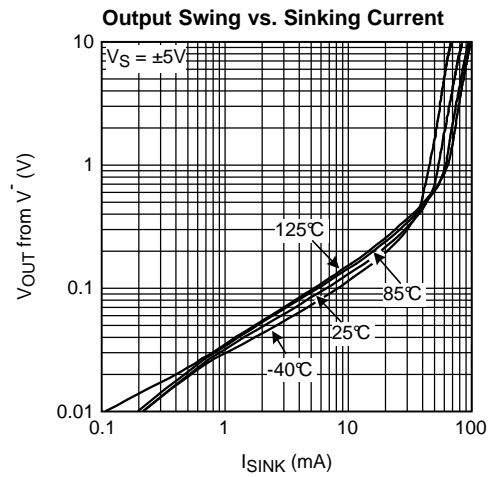


Figure 9.

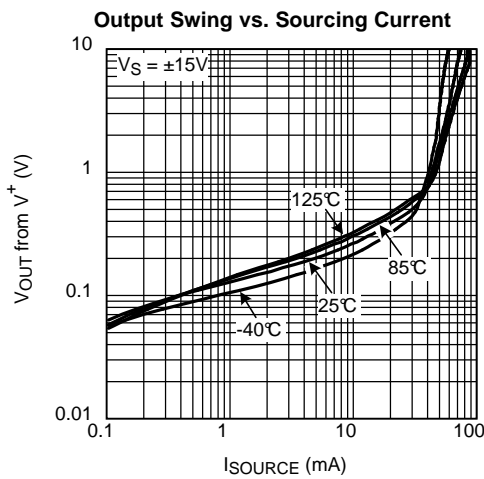


Figure 10.

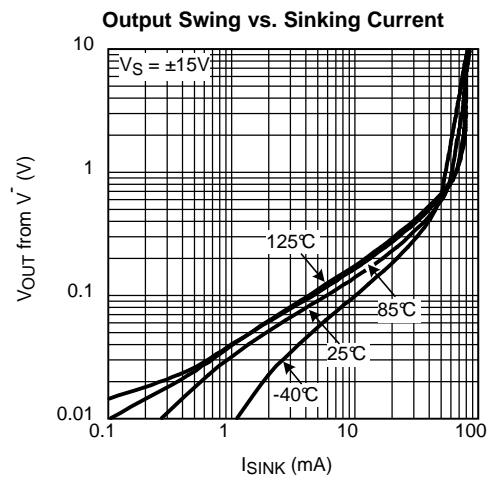


Figure 11.

Typical Performance Characteristics (continued)

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

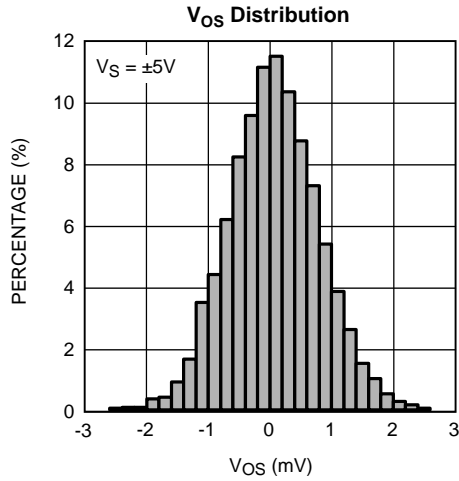


Figure 12.

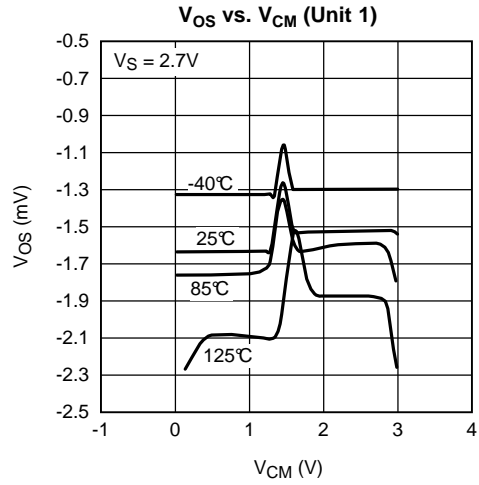


Figure 13.

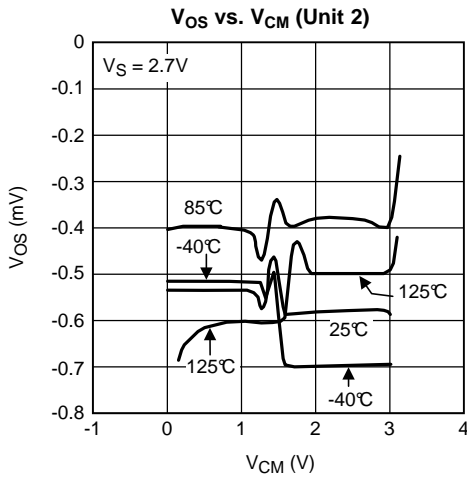


Figure 14.

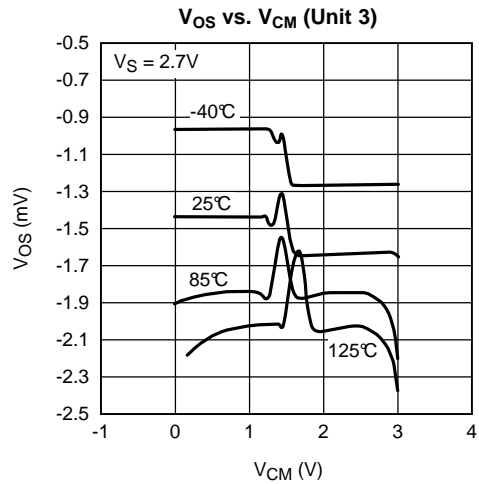


Figure 15.

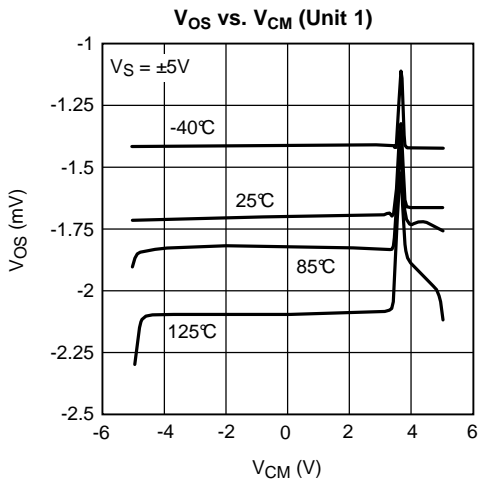


Figure 16.

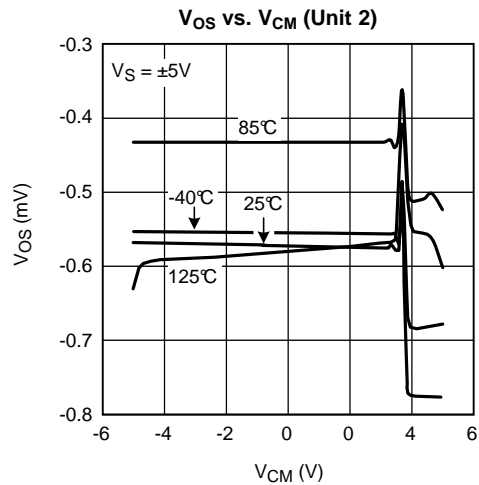


Figure 17.

Typical Performance Characteristics (continued)

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

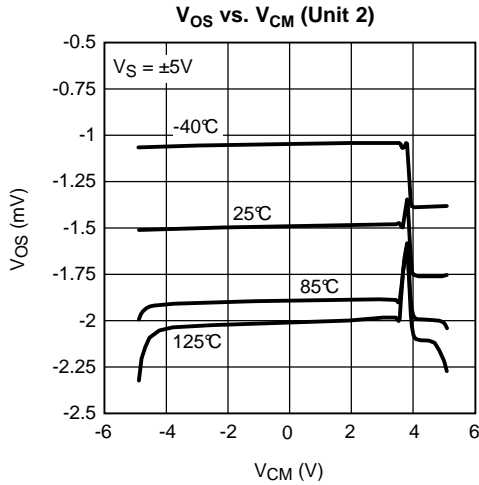


Figure 18.

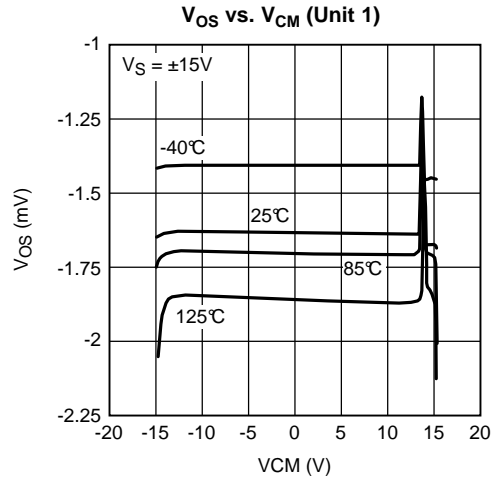


Figure 19.

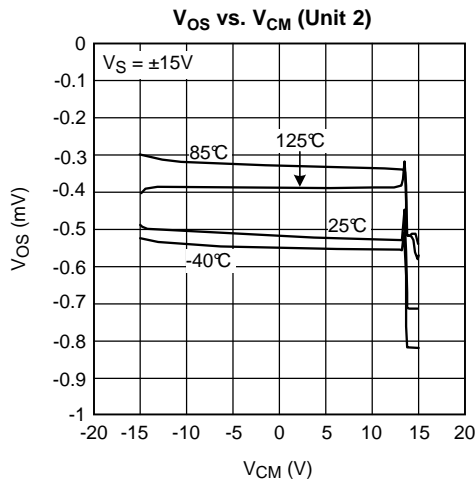


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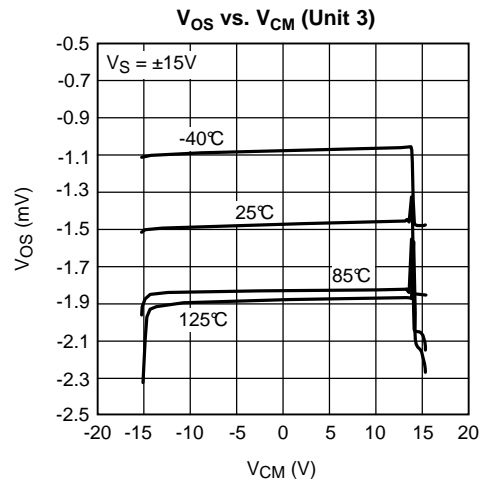


Figure 21.

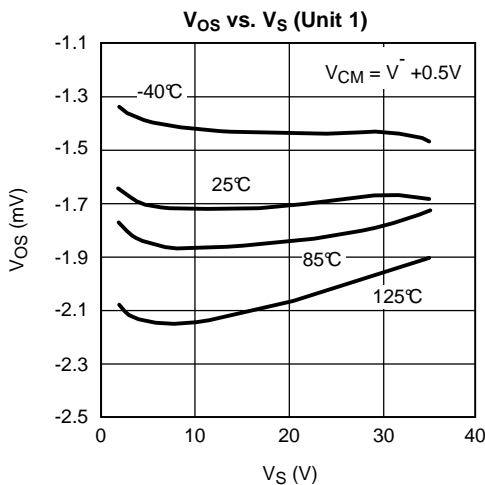


Figure 22.

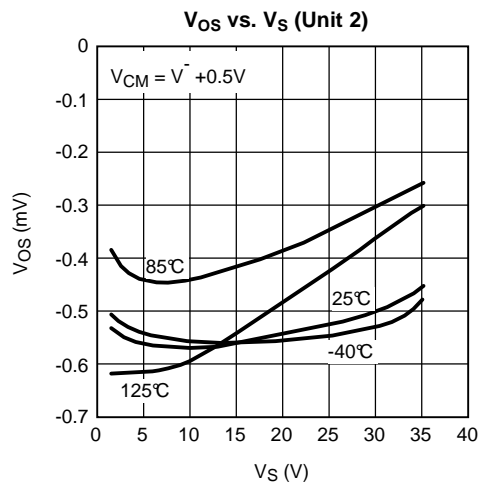


Figure 23.

Typical Performance Characteristics (continued)

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

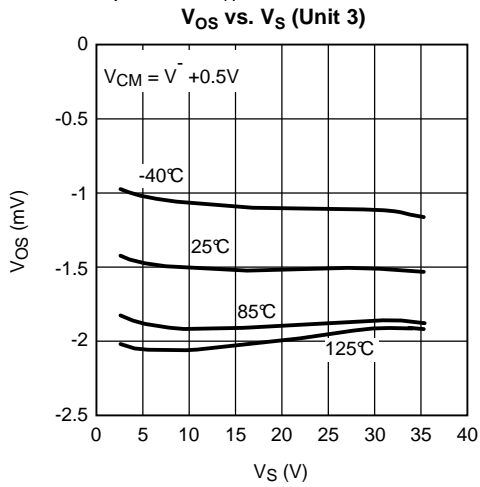


Figure 24.

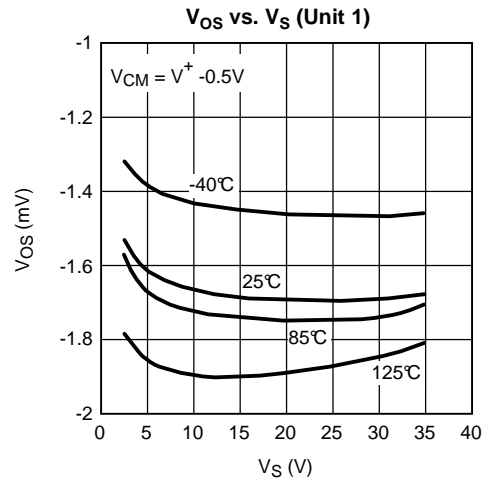


Figure 25.

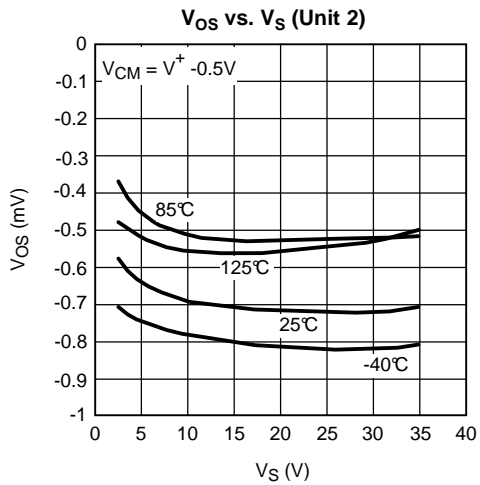


Figure 26.

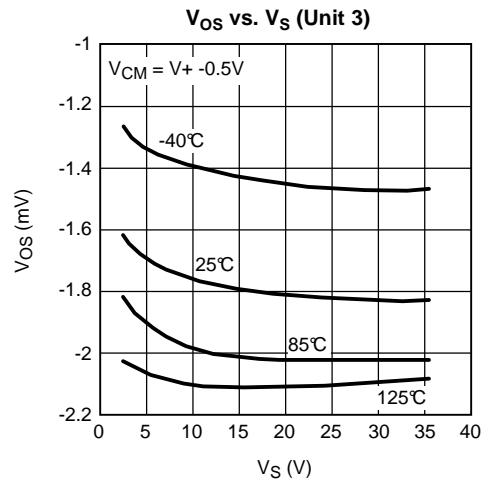


Figure 27.

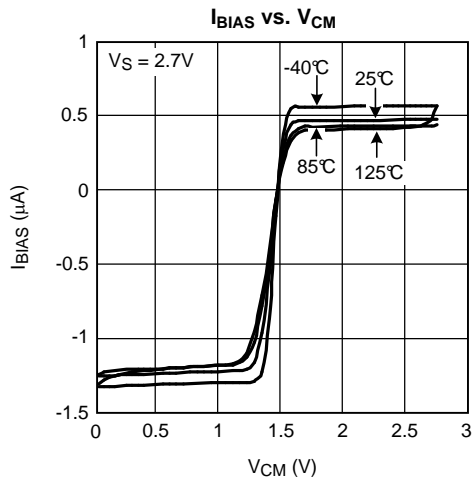


Figure 28.

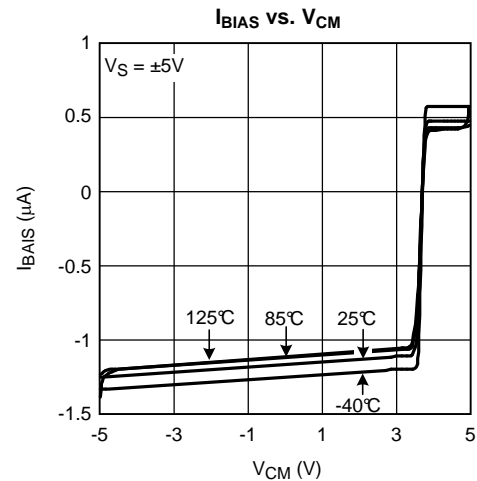


Figure 29.

**Typical Performance Characteristics (continued)**

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

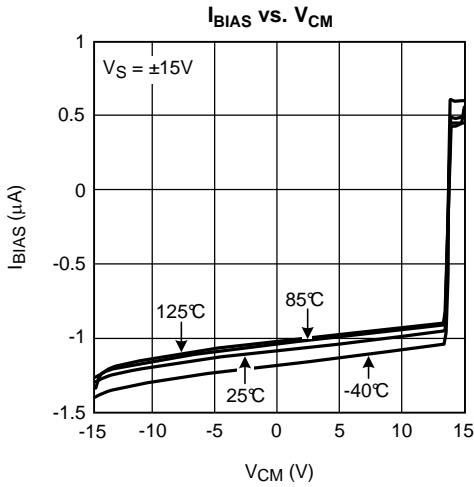


Figure 30.

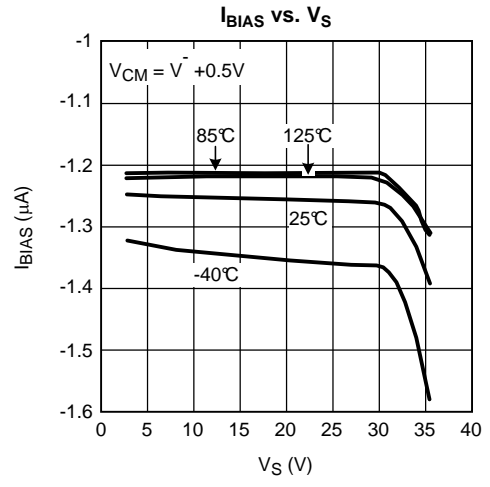


Figure 31.

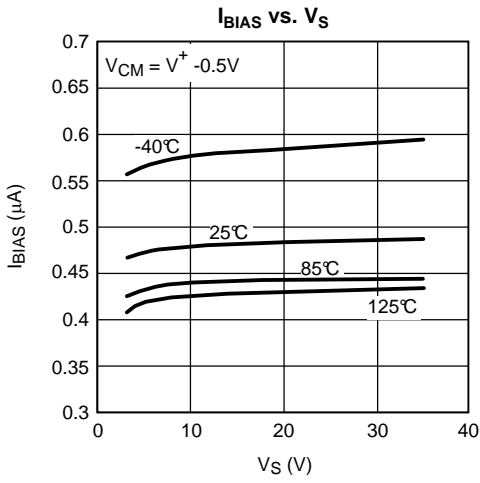


Figure 32.

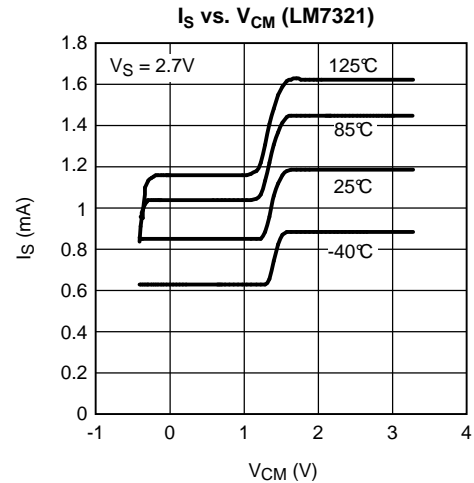


Figure 33.

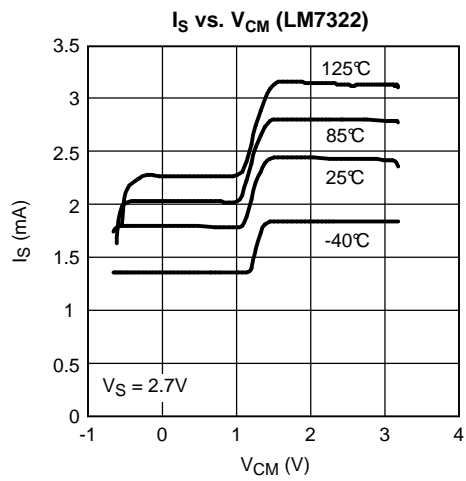


Figure 34.

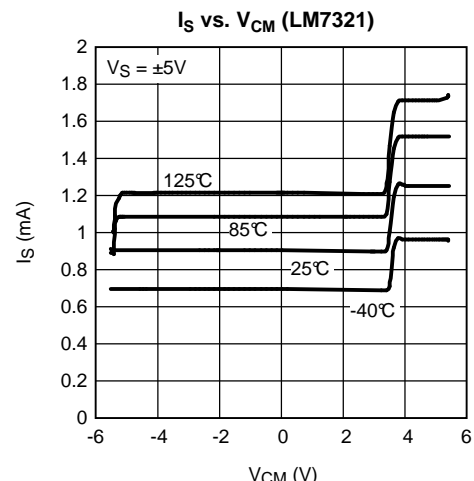


Figure 35.

Typical Performance Characteristics (continued)

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

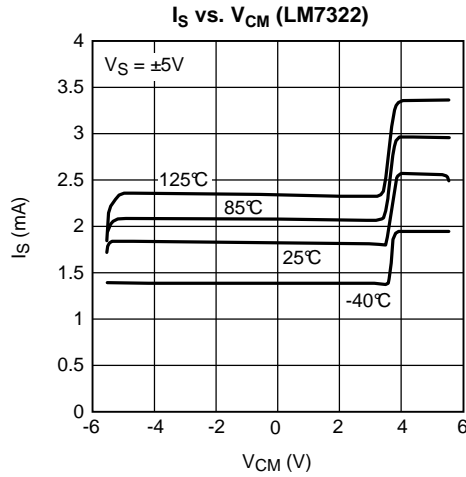


Figure 36.

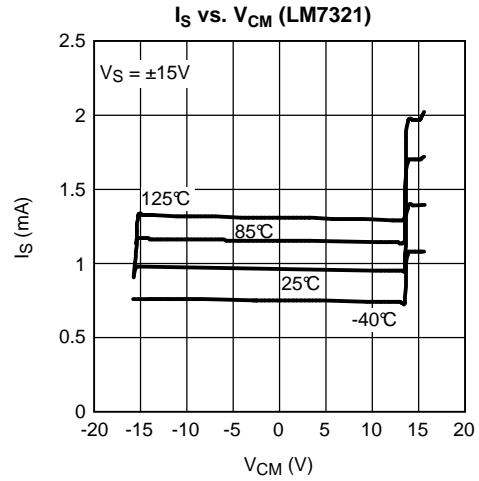


Figure 37.

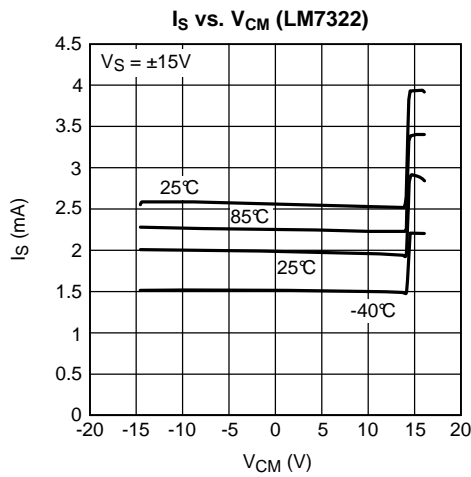


Figure 38.

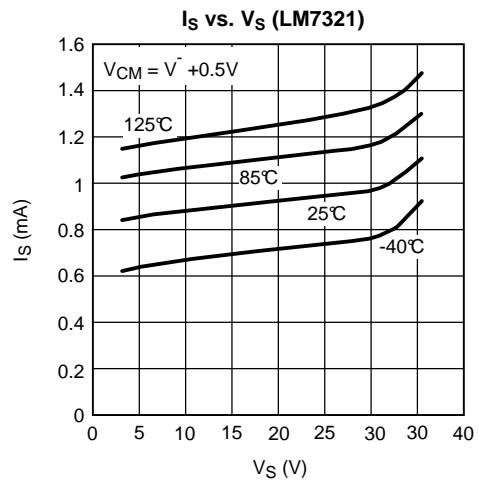


Figure 39.

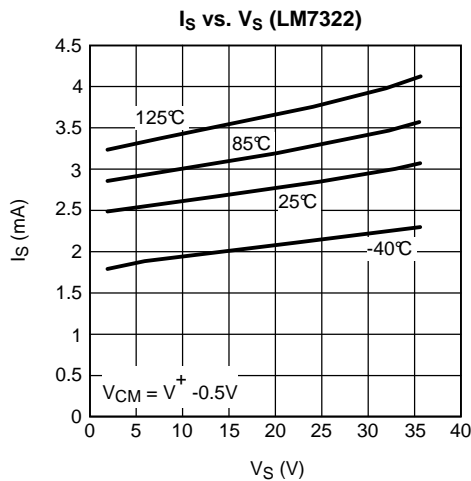


Figure 40.

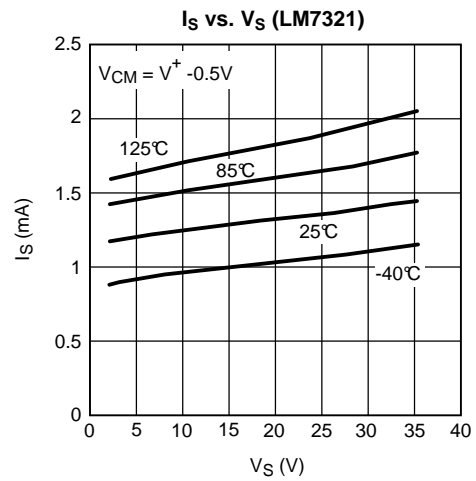


Figure 41.

**Typical Performance Characteristics (continued)**

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

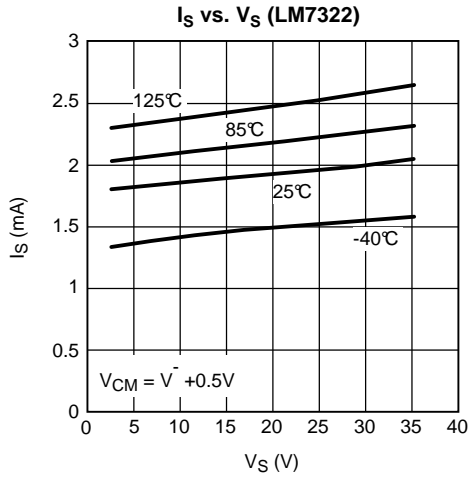


Figure 42.

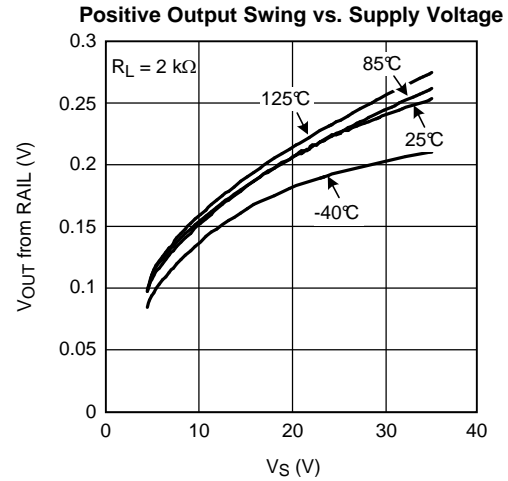


Figure 43.

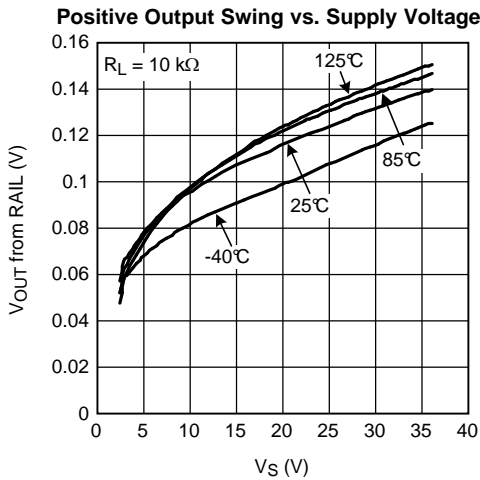


Figure 44.

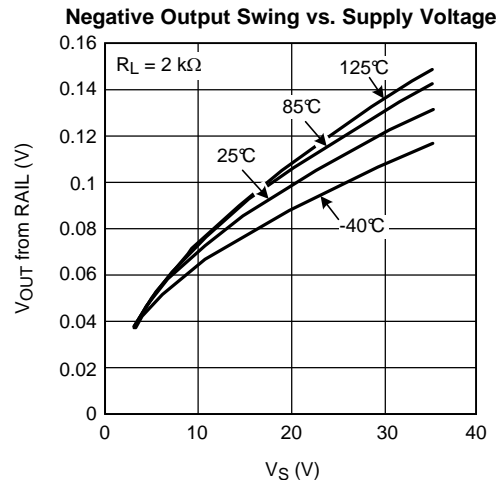


Figure 45.

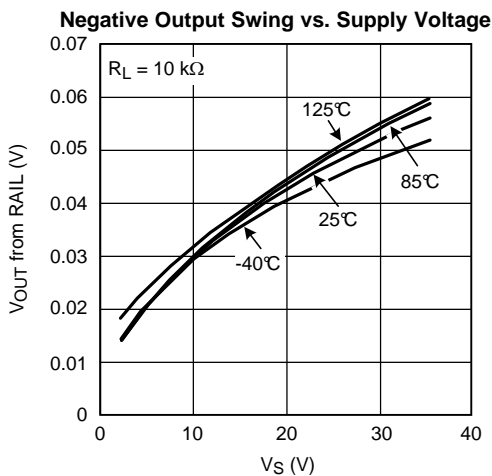


Figure 46.

**Open Loop Frequency Response with Various Capacitive Load**

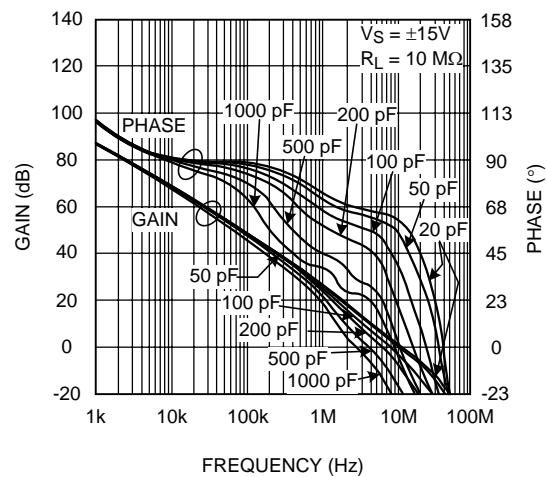


Figure 47.

Typical Performance Characteristics (continued)

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

Open Loop Frequency Response with Various Resistive Load

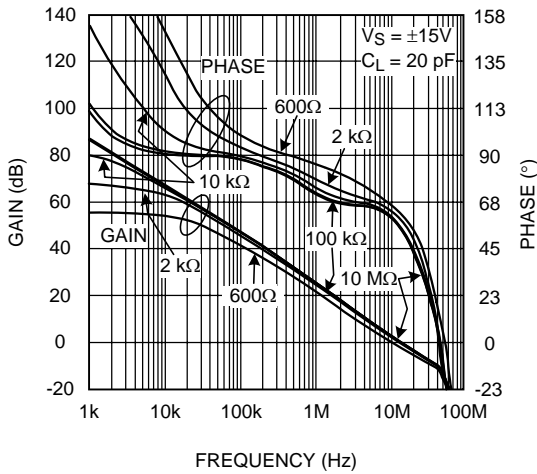


Figure 48.

Open Loop Frequency Response with Various Supply Voltage

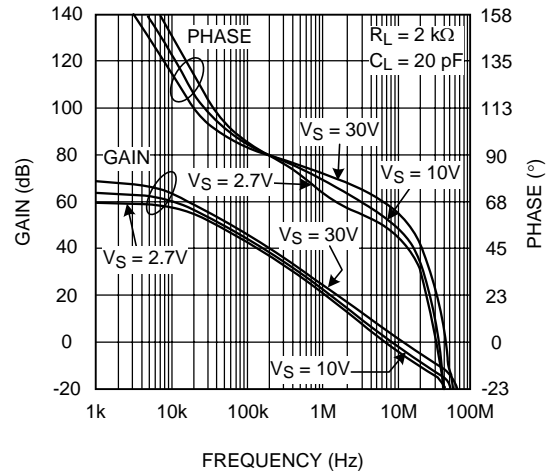


Figure 49.

Phase Margin vs. Capacitive Load

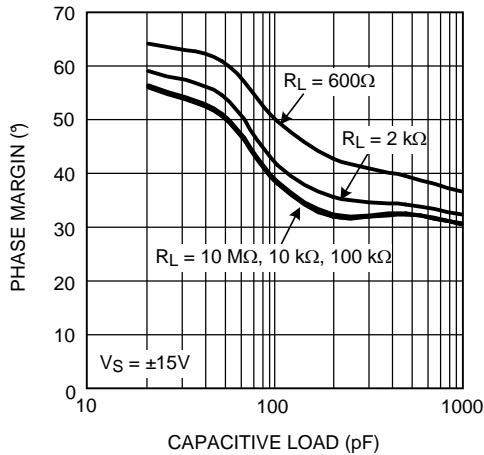


Figure 50.

CMRR vs. Frequency

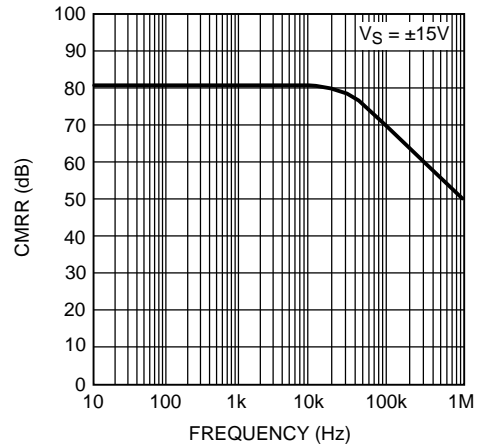


Figure 51.

+PSRR vs. Frequency

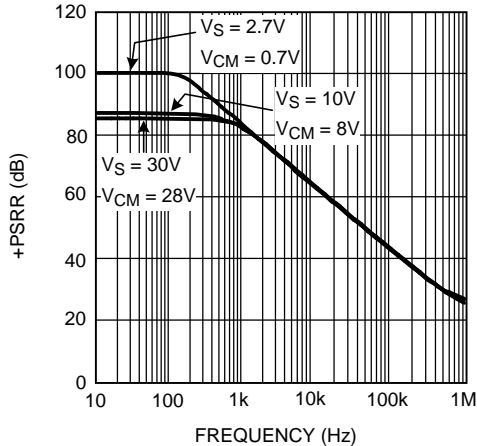


Figure 52.

-PSRR vs. Frequency

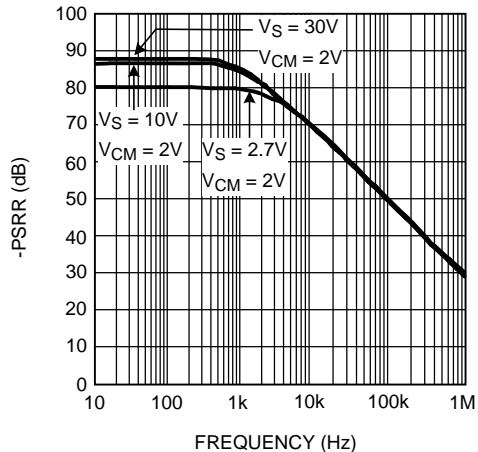


Figure 53.

**Typical Performance Characteristics (continued)**

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

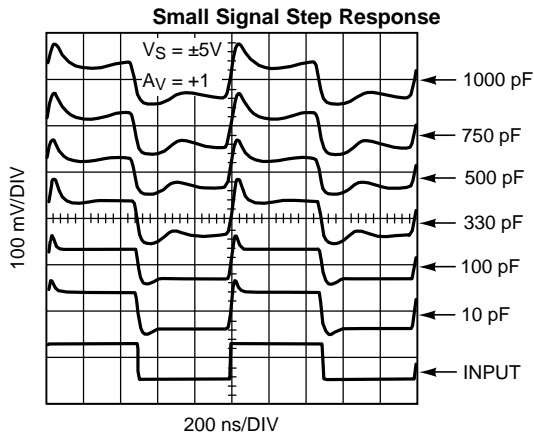


Figure 54.

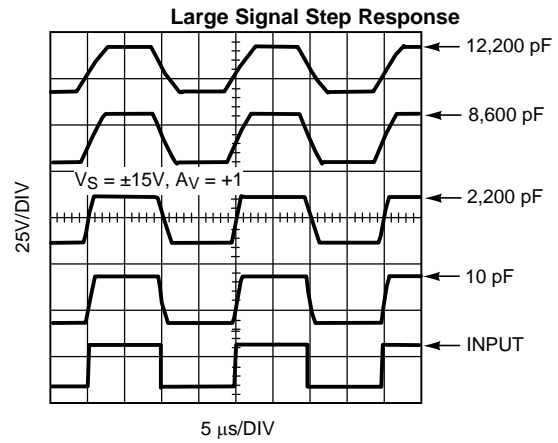


Figure 55.

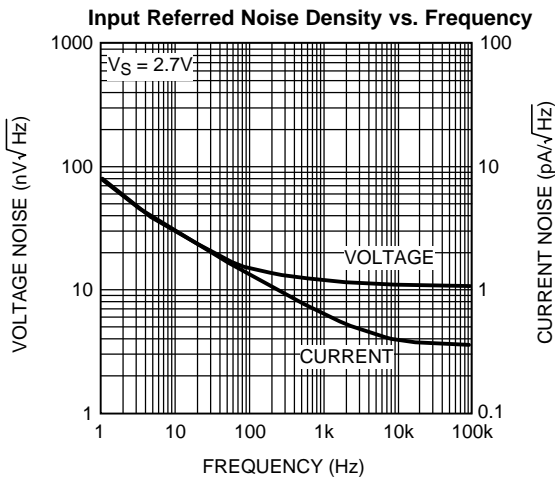


Figure 56.

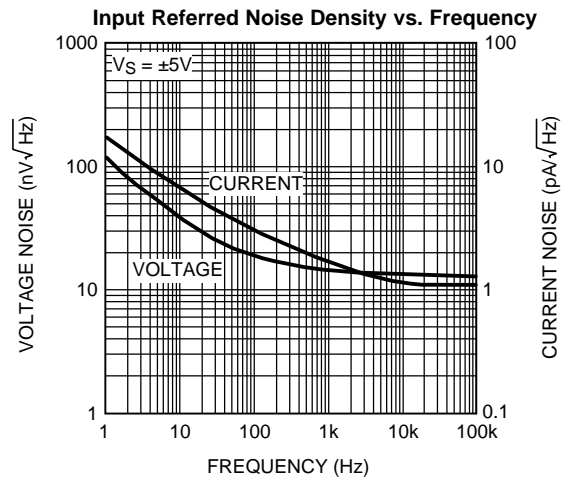


Figure 57.

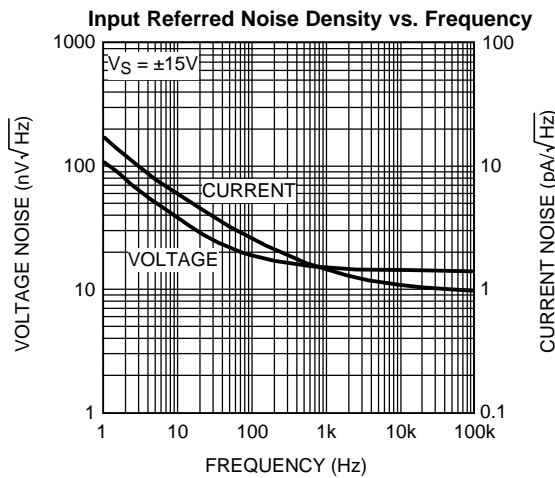


Figure 58.

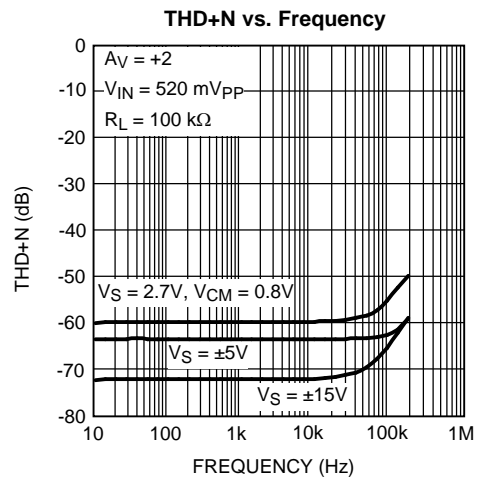


Figure 59.

### Typical Performance Characteristics (continued)

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ .

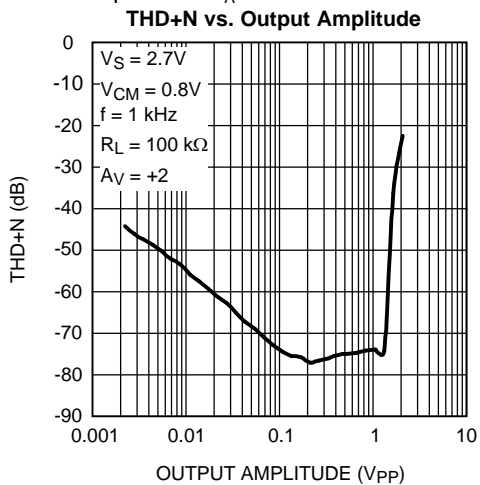


Figure 60.

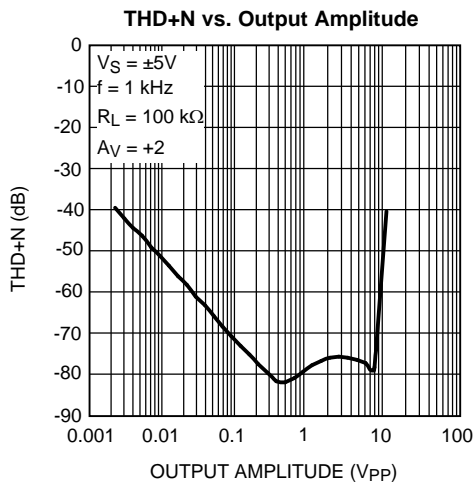


Figure 61.

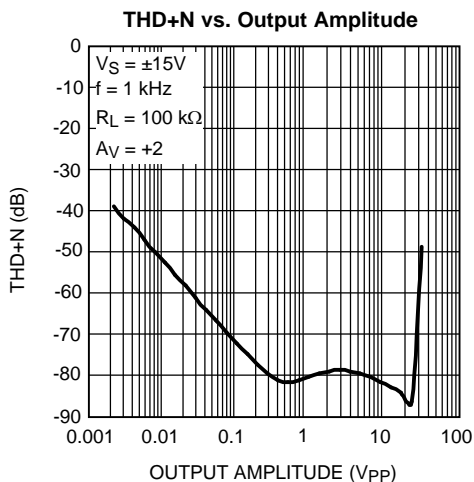


Figure 62.

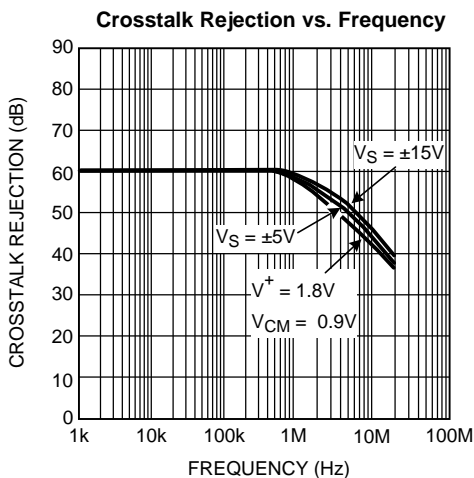
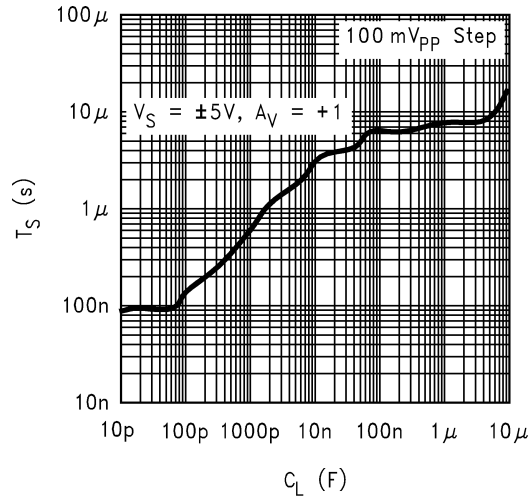


Figure 63.

## APPLICATION INFORMATION

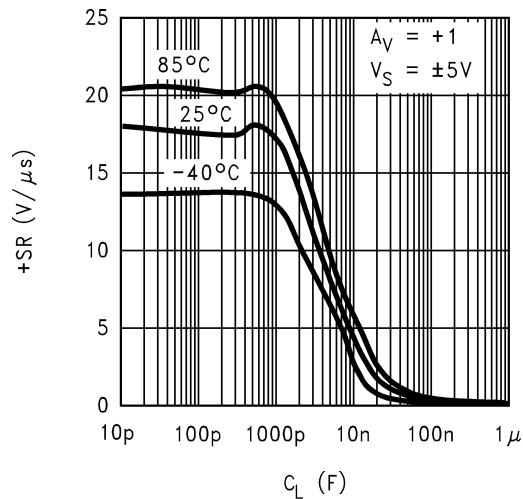
### DRIVING CAPACITIVE LOADS

The LM7321/LM7321Q/LM7322/LM7322Q are specifically designed to drive unlimited capacitive loads without oscillations as shown in [Figure 64](#).



**Figure 64. ±5% Settling Time vs. Capacitive Load**

In addition, the output current handling capability of the device allows for good slewing characteristics even with large capacitive loads as shown in [Figure 65](#) and [Figure 66](#).



**Figure 65. +SR vs. Capacitive Load**

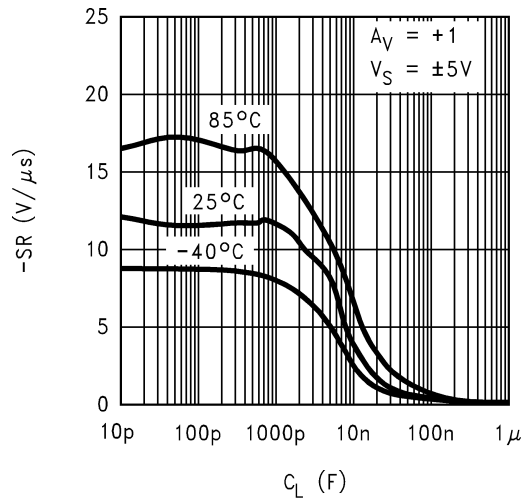


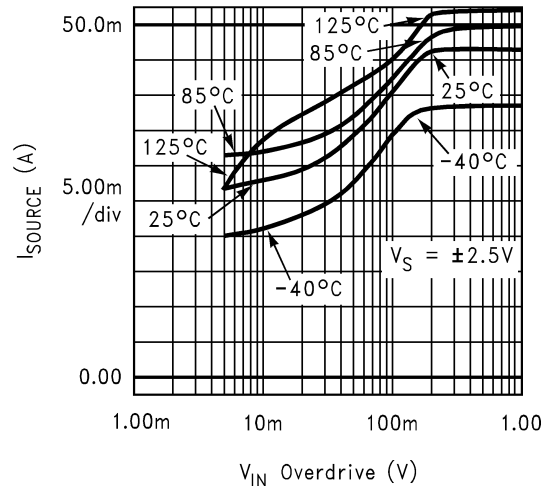
Figure 66. -SR vs. Capacitive Load

The combination of these features is ideal for applications such as TFT flat panel buffers, A/D converter input amplifiers, etc.

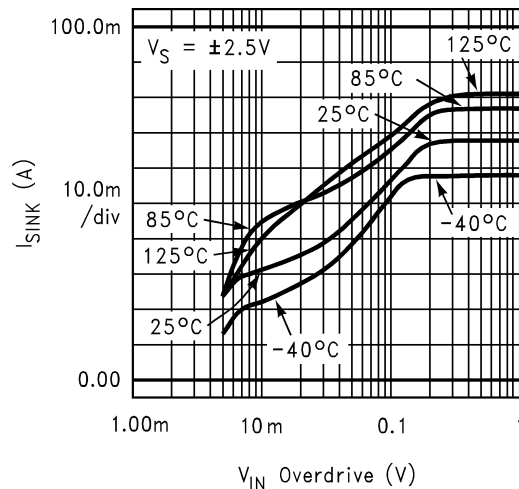
However, as in most op amps, addition of a series isolation resistor between the op amp and the capacitive load improves the settling and overshoot performance.

Output current drive is an important parameter when driving capacitive loads. This parameter will determine how fast the output voltage can change. Referring to the Slew Rate vs. Capacitive Load Plots ([Typical Performance Characteristics](#) section), two distinct regions can be identified. Below about 10,000 pF, the output Slew Rate is solely determined by the op amp's compensation capacitor value and available current into that capacitor. Beyond 10 nF, the Slew Rate is determined by the op amp's available output current. Note that because of the lower output sourcing current compared to the sinking one, the Slew Rate limit under heavy capacitive loading is determined by the positive transitions. An estimate of positive and negative slew rates for loads larger than 100 nF can be made by dividing the short circuit current value by the capacitor.

For the LM7321/LM7321Q/LM7322/LM7322Q, the available output current increases with the input overdrive. Referring to [Figure 67](#) and [Figure 68](#), Output Short Circuit Current vs. Input Overdrive, it can be seen that both sourcing and sinking short circuit current increase as input overdrive increases. In a closed loop amplifier configuration, during transient conditions while the fed back output has not quite caught up with the input, there will be an overdrive imposed on the input allowing more output current than would normally be available under steady state condition. Because of this feature, the op amp's output stage quiescent current can be kept to a minimum, thereby reducing power consumption, while enabling the device to deliver large output current when the need arises (such as during transients).



**Figure 67. Output Short Circuit Sourcing Current vs. Input Overdrive**



**Figure 68. Output Short Circuit Sinking Current vs. Input Overdrive**

Figure 69 shows the output voltage, output current, and the resulting input overdrive with the device set for  $A_V = +1$  and the input tied to a 1 V<sub>PP</sub> step function driving a 47 nF capacitor. As can be seen, during the output transition, the input overdrive reaches 1V peak and is more than enough to cause the output current to increase to its maximum value (see Figure 67 and Figure 68 plots). Note that because of the larger output sinking current compared to the sourcing one, the output negative transition is faster than the positive one.

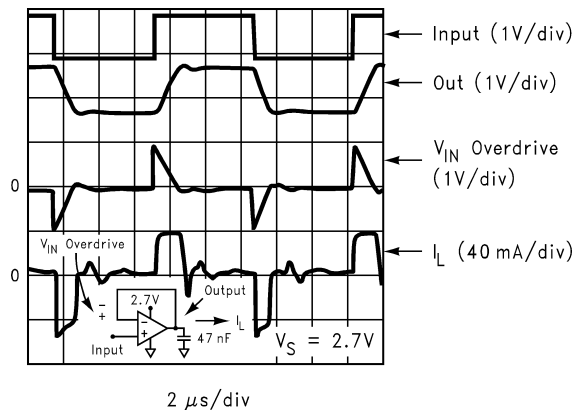


Figure 69. Buffer Amplifier Scope Photo

### ESTIMATING THE OUTPUT VOLTAGE SWING

It is important to keep in mind that the steady state output current will be less than the current available when there is an input overdrive present. For steady state conditions, the Output Voltage vs. Output Current plot (Typical Performance Characteristics section) can be used to predict the output swing. Figure 70 and Figure 71 show this performance along with several load lines corresponding to loads tied between the output and ground. In each cases, the intersection of the device plot at the appropriate temperature with the load line would be the typical output swing possible for that load. For example, a 1 kΩ load can accommodate an output swing to within 250 mV of  $V^-$  and to 330 mV of  $V^+$  ( $V_S = \pm 15V$ ) corresponding to a typical 29.3 V<sub>PP</sub> unclipped swing.

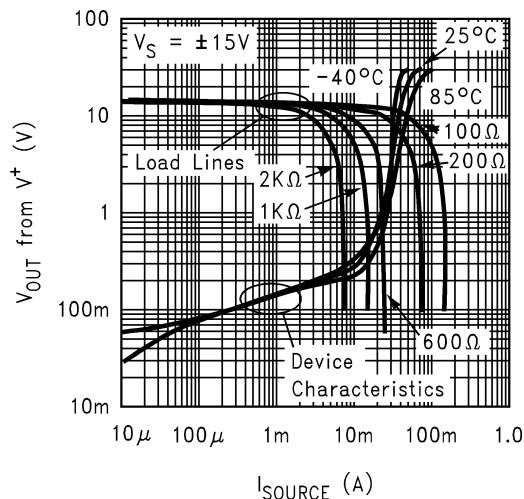
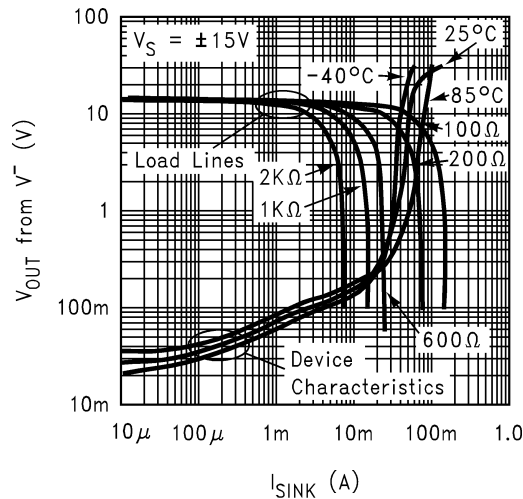


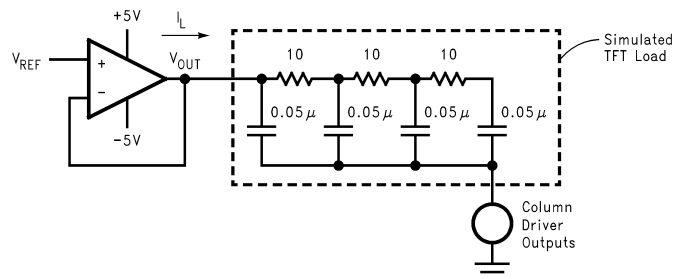
Figure 70. Output Sourcing Characteristics with Load Lines



**Figure 71. Output Sinking Characteristics with Load Lines**

### SETTLING TIME WITH LARGE CAPACITIVE LOADS

Figure 72 below shows a typical application where the LM7321/LM7321Q/LM7322/LM7322Q is used as a buffer amplifier for the  $V_{COM}$  signal employed in a TFT LCD flat panel:



**Figure 72.  $V_{COM}$  Driver Application Schematic**

Figure 73 shows the time domain response of the amplifier when used as a  $V_{COM}$  buffer/driver with  $V_{REF}$  at ground. In this application, the op amp loop will try and maintain its output voltage based on the voltage on its non-inverting input ( $V_{REF}$ ) despite the current injected into the TFT simulated load. As long as this load current is within the range tolerable by the LM7321/LM7321Q/LM7322/LM7322Q (45 mA sourcing and 65 mA sinking for  $\pm 5V$  supplies), the output will settle to its final value within less than 2  $\mu s$ .

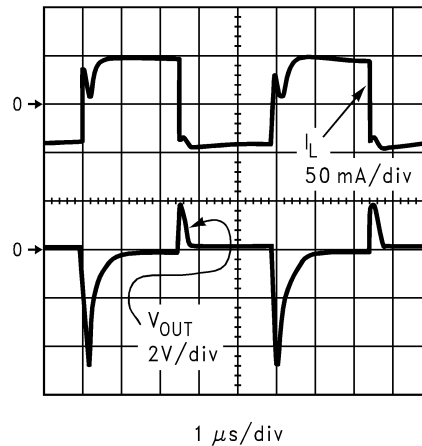


Figure 73.  $V_{COM}$  Driver Performance Scope Photo

### OUTPUT SHORT CIRCUIT CURRENT AND DISSIPATION ISSUES

The LM7321/LM7321Q/LM7322/LM7322Q output stage is designed for maximum output current capability. Even though momentary output shorts to ground and either supply can be tolerated at all operating voltages, longer lasting short conditions can cause the junction temperature to rise beyond the absolute maximum rating of the device, especially at higher supply voltage conditions. Below supply voltage of 6V, the output short circuit condition can be tolerated indefinitely.

With the op amp tied to a load, the device power dissipation consists of the quiescent power due to the supply current flow into the device, in addition to power dissipation due to the load current. The load portion of the power itself could include an average value (due to a DC load current) and an AC component. DC load current would flow if there is an output voltage offset, or the output AC average current is non-zero, or if the op amp operates in a single supply application where the output is maintained somewhere in the range of linear operation.

Therefore:

$P_{TOTAL} = P_Q + P_{DC} + P_{AC}$	
$P_Q = I_S \cdot V_S$	Op Amp Quiescent Power Dissipation
$P_{DC} = I_O \cdot (V_r - V_o)$	DC Load Power
$P_{AC} = \text{See Table 1}$	AC Load Power

where:

$I_S$ : Supply Current

$V_S$ : Total Supply Voltage ( $V^+ - V^-$ )

$V_O$ : Average Output Voltage

$V_r$ :  $V^+$  for sourcing and  $V^-$  for sinking current

Table 1 shows the maximum AC component of the load power dissipated by the op amp for standard Sinusoidal, Triangular, and Square Waveforms:

Table 1. Normalized AC Power Dissipated in the Output Stage for Standard Waveforms

	$P_{AC} (W \cdot \Omega / V^2)$	
	Triangular	Square
Sinusoidal	$46.9 \times 10^{-3}$	$62.5 \times 10^{-3}$
	$50.7 \times 10^{-3}$	

The table entries are normalized to  $V_S^2/R_L$ . To figure out the AC load current component of power dissipation, simply multiply the table entry corresponding to the output waveform by the factor  $V_S^2/R_L$ . For example, with  $\pm 12V$  supplies, a  $600\Omega$  load, and triangular waveform power dissipation in the output stage is calculated as:

$$P_{AC} = (46.9 \times 10^{-3}) \cdot (242/600) = 45.0 \text{ mW} \quad (1)$$

The maximum power dissipation allowed at a certain temperature is a function of maximum die junction temperature ( $T_{J(MAX)}$ ) allowed, ambient temperature  $T_A$ , and package thermal resistance from junction to ambient,  $\theta_{JA}$ .

$$P_{D(MAX)} = \frac{T_{J(MAX)} - T_A}{\theta_{JA}} \quad (2)$$

For the LM7321/LM7321Q/LM7322/LM7322Q, the maximum junction temperature allowed is  $150^\circ\text{C}$  at which no power dissipation is allowed. The power capability at  $25^\circ\text{C}$  is given by the following calculations:

For VSSOP package:

$$P_{D(MAX)} = \frac{150^\circ\text{C} - 25^\circ\text{C}}{235^\circ\text{C/W}} = 0.53\text{W} \quad (3)$$

For SOIC package:

$$P_{D(MAX)} = \frac{150^\circ\text{C} - 25^\circ\text{C}}{165^\circ\text{C/W}} = 0.76\text{W} \quad (4)$$

Similarly, the power capability at  $125^\circ\text{C}$  is given by:

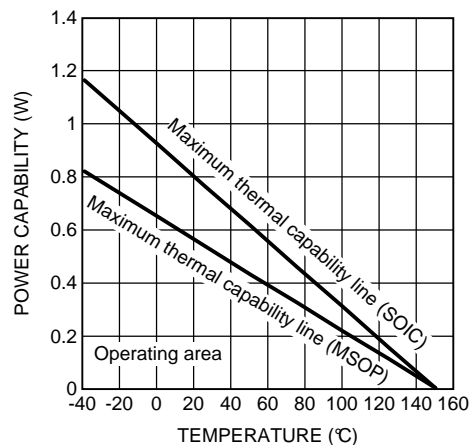
For VSSOP package:

$$P_{D(MAX)} = \frac{150^\circ\text{C} - 125^\circ\text{C}}{235^\circ\text{C/W}} = 0.11\text{W} \quad (5)$$

For SOIC package:

$$P_{D(MAX)} = \frac{150^\circ\text{C} - 125^\circ\text{C}}{165^\circ\text{C/W}} = 0.15\text{W} \quad (6)$$

Figure 74 shows the power capability vs. temperature for VSSOP and SOIC packages. The area under the maximum thermal capability line is the operating area for the device. When the device works in the operating area where  $P_{TOTAL}$  is less than  $P_{D(MAX)}$ , the device junction temperature will remain below  $150^\circ\text{C}$ . If the intersection of ambient temperature and package power is above the maximum thermal capability line, the junction temperature will exceed  $150^\circ\text{C}$  and this should be strictly prohibited.



**Figure 74. Power Capability vs. Temperature**

When high power is required and ambient temperature can't be reduced, providing air flow is an effective approach to reduce thermal resistance therefore to improve power capability.

## Other Application Hints

The use of supply decoupling is mandatory in most applications. As with most relatively high speed/high output current Op Amps, best results are achieved when each supply line is decoupled with two capacitors; a small value ceramic capacitor ( $\sim 0.01 \mu\text{F}$ ) placed very close to the supply lead in addition to a large value Tantalum or Aluminum ( $> 4.7 \mu\text{F}$ ). The large capacitor can be shared by more than one device if necessary. The small ceramic capacitor maintains low supply impedance at high frequencies while the large capacitor will act as the charge "bucket" for fast load current spikes at the op amp output. The combination of these capacitors will provide supply decoupling and will help keep the op amp oscillation free under any load.

## SIMILAR HIGH OUTPUT DEVICES

The LM7332 is a dual rail-to-rail amplifier with a slightly lower GBW capable of sinking and sourcing 100 mA. It is available in SOIC and VSSOP packages.

The LM4562 is dual op amp with very low noise and 0.7 mV voltage offset.

The LME49870 and LME49860 are single and dual low noise amplifiers that can work from  $\pm 22$  volt supplies.

## OTHER HIGH PERFORMANCE SOT-23 AMPLIERS

The LM7341 is a 4 MHz rail-to-rail input and output part that requires only 0.6 mA to operate, and can drive unlimited capacitive load. It has a voltage gain of 97 dB, a CMRR of 93 dB, and a PSRR of 104 dB.

The LM6211 is a 20 MHz part with CMOS input, which runs on  $\pm 12$  volt or 24 volt single supplies. It has rail-to-rail output and low noise.

The LM7121 has a gain bandwidth of 235 MHz.

Detailed information on these parts can be found at [www.ti.com](http://www.ti.com).

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## REVISION HISTORY

Changes from Revision C (March 2013) to Revision D	Page
• Changed layout of National Data Sheet to TI format .....	<a href="#">25</a>

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**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
LM7321MA/NOPB	ACTIVE	SOIC	D	8	95	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	LM732 1MA	<a href="#">Samples</a>
LM7321MAX/NOPB	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	LM732 1MA	<a href="#">Samples</a>
LM7321MF/NOPB	ACTIVE	SOT-23	DBV	5	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	AU4A	<a href="#">Samples</a>
LM7321MFE/NOPB	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	AU4A	<a href="#">Samples</a>
LM7321MFX/NOPB	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	AU4A	<a href="#">Samples</a>
LM7321QMF/NOPB	ACTIVE	SOT-23	DBV	5	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	AR8A	<a href="#">Samples</a>
LM7321QMFE/NOPB	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	AR8A	<a href="#">Samples</a>
LM7321QMFX/NOPB	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	AR8A	<a href="#">Samples</a>
LM7322MA/NOPB	ACTIVE	SOIC	D	8	95	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	LM732 2MA	<a href="#">Samples</a>
LM7322MAX/NOPB	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	LM732 2MA	<a href="#">Samples</a>
LM7322MM/NOPB	ACTIVE	VSSOP	DGK	8	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	AZ4A	<a href="#">Samples</a>
LM7322MME/NOPB	ACTIVE	VSSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	AZ4A	<a href="#">Samples</a>
LM7322MMX/NOPB	ACTIVE	VSSOP	DGK	8	3500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	AZ4A	<a href="#">Samples</a>
LM7322QMA/NOPB	ACTIVE	SOIC	D	8	95	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	LM732 2QMA	<a href="#">Samples</a>
LM7322QMAX/NOPB	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	LM732 2QMA	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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**OTHER QUALIFIED VERSIONS OF LM7321, LM7321-Q1, LM7322, LM7322-Q1 :**

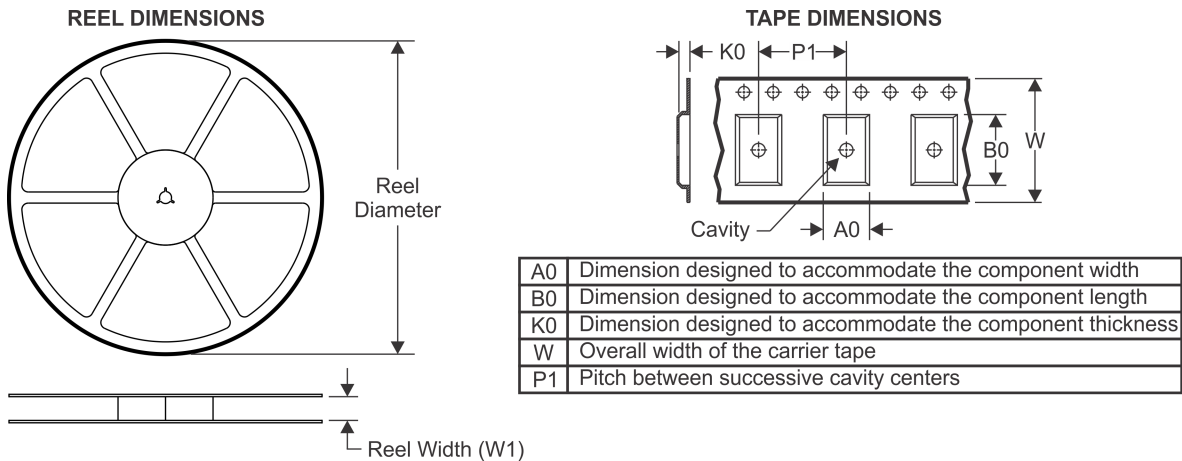
● Catalog: [LM7321](#), [LM7322](#)

● Automotive: [LM7321-Q1](#), [LM7322-Q1](#)

**NOTE: Qualified Version Definitions:**

● Catalog - TI's standard catalog product

● Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM7321MAX/NOPB	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LM7321MF/NOPB	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LM7321MFE/NOPB	SOT-23	DBV	5	250	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LM7321MFX/NOPB	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LM7321QMF/NOPB	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LM7321QMFE/NOPB	SOT-23	DBV	5	250	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LM7321QMFN/NOPB	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LM7322MAX/NOPB	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LM7322MM/NOPB	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LM7322MME/NOPB	VSSOP	DGK	8	250	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LM7322MMX/NOPB	VSSOP	DGK	8	3500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LM7322QMAX/NOPB	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1

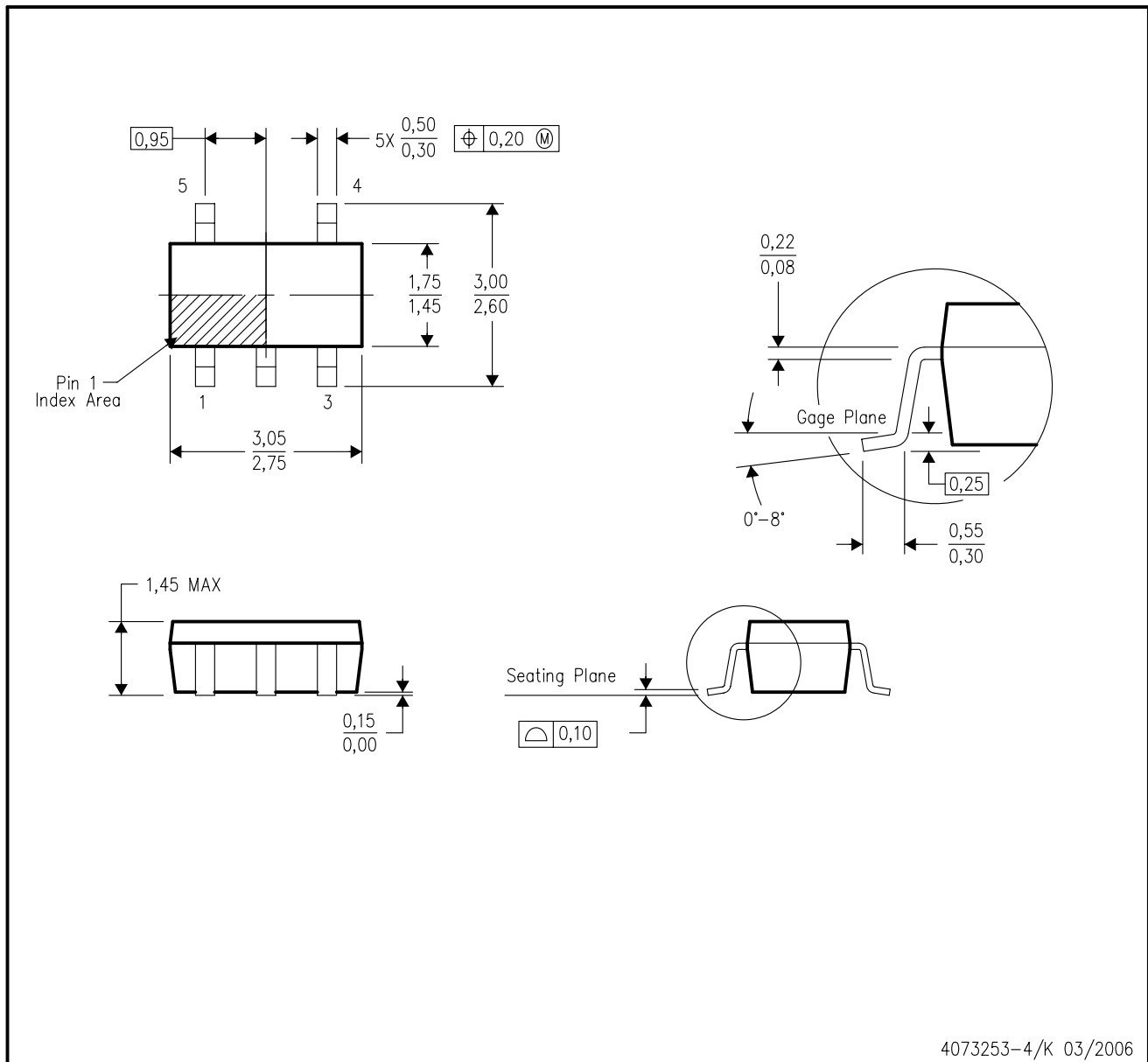
**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM7321MAX/NOPB	SOIC	D	8	2500	367.0	367.0	35.0
LM7321MF/NOPB	SOT-23	DBV	5	1000	210.0	185.0	35.0
LM7321MFE/NOPB	SOT-23	DBV	5	250	210.0	185.0	35.0
LM7321MFX/NOPB	SOT-23	DBV	5	3000	210.0	185.0	35.0
LM7321QMF/NOPB	SOT-23	DBV	5	1000	210.0	185.0	35.0
LM7321QMFE/NOPB	SOT-23	DBV	5	250	210.0	185.0	35.0
LM7321QMFX/NOPB	SOT-23	DBV	5	3000	210.0	185.0	35.0
LM7322MAX/NOPB	SOIC	D	8	2500	367.0	367.0	35.0
LM7322MM/NOPB	VSSOP	DGK	8	1000	210.0	185.0	35.0
LM7322MME/NOPB	VSSOP	DGK	8	250	210.0	185.0	35.0
LM7322MMX/NOPB	VSSOP	DGK	8	3500	367.0	367.0	35.0
LM7322QMAX/NOPB	SOIC	D	8	2500	367.0	367.0	35.0

DBV (R-PDSO-G5)

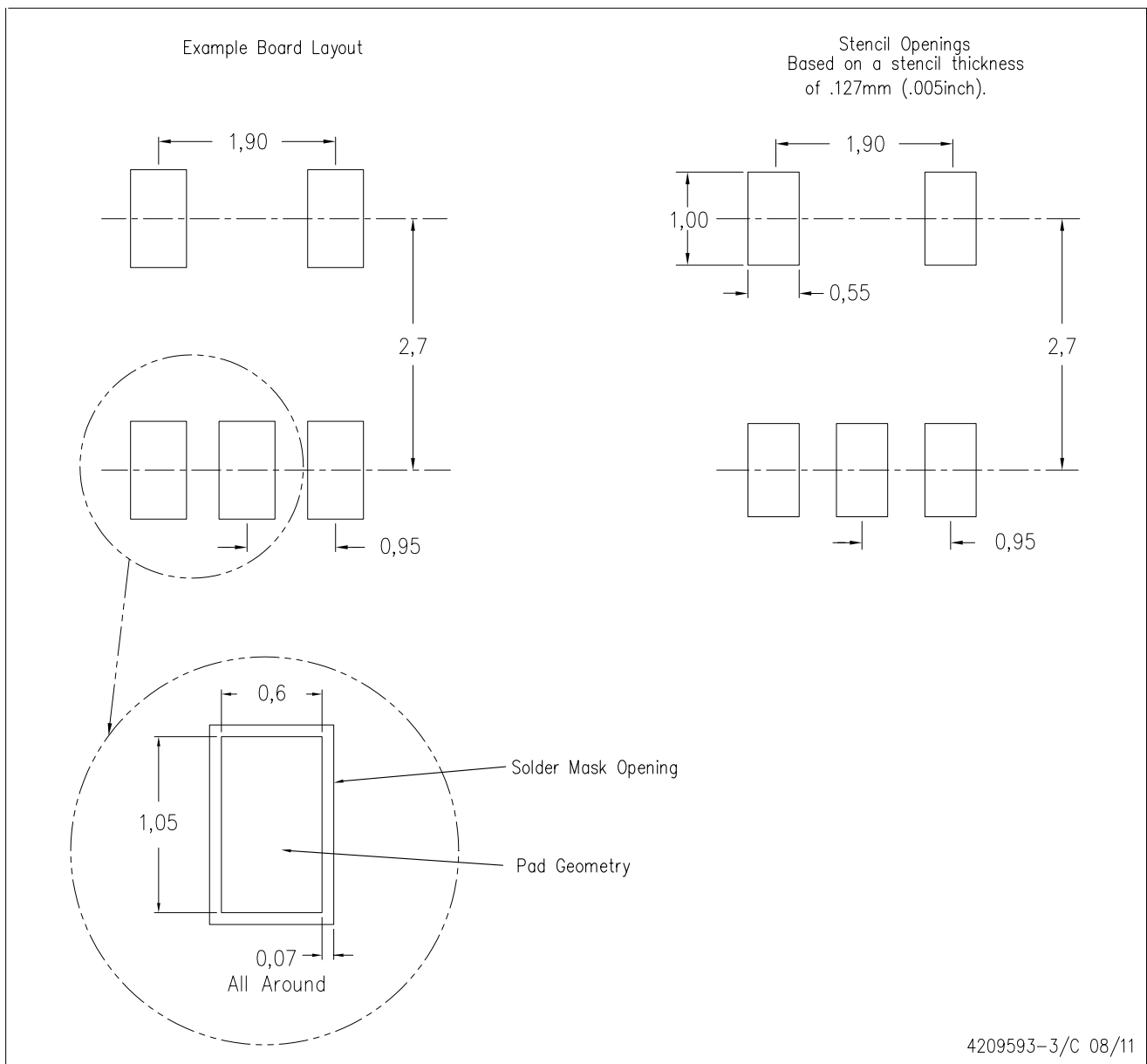
PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
  - D. Falls within JEDEC MO-178 Variation AA.

DBV (R-PDSO-G5)

PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
  - D. Publication IPC-7351 is recommended for alternate designs.
  - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.





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Clocks and Timers	<a href="http://www.ti.com/clocks">www.ti.com/clocks</a>
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Security	<a href="http://www.ti.com/security">www.ti.com/security</a>
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