

FEATURES

- Fully differential signal path, also used with single-sided signals**
- Inputs from 0.3 mV to 1 V rms, rail-to-rail outputs**
- Differential $R_{IN} = 1\text{ k}\Omega$; R_{OUT} (each output) $75\ \Omega$**
- Automatic offset compensation (optional)**
- Linear-in-dB and linear-in-magnitude gain modes**
- 0 dB to 50 dB, for $0\text{ V} < V_{DBS} < 1.5\text{ V}$ (30 mV/dB)**
- Inverted gain mode: 50 dB to 0 dB at -30 mV/dB**
- $\times 0.03$ to $\times 10$ nominal gain for $15\text{ mV} < V_{MAG} < 5\text{ V}$**
- Constant bandwidth: 150 MHz at all gains**
- Low noise: 5 nV/ $\sqrt{\text{Hz}}$ typical at maximum gain**
- Low distortion: $\leq -62\text{ dBc}$ typical**
- Low power: 20 mA typical at V_S of 2.7 V to 6 V**
- Available in a space-saving, 3 mm \times 3 mm LFCSP package**

APPLICATIONS

- Pre-ADC signal conditioning**
- 75 Ω cable driving adjust**
- AGC amplifiers**

GENERAL DESCRIPTION

The AD8330¹ is a wideband variable gain amplifier for applications requiring a fully differential signal path, low noise, well-defined gain, and moderately low distortion, from dc to 150 MHz. The input pins can also be driven from a single-ended source. The peak differential input is $\pm 2\text{ V}$, allowing sine wave operation at 1 V rms with generous headroom. The output pins can drive single-sided loads essentially rail-to-rail. The differential output resistance is 150 Ω . The output swing is a linear function of the voltage applied to the VMAG pin that internally defaults to 0.5 V, providing a peak output of $\pm 2\text{ V}$. This can be raised to 10 V p-p, limited by the supply voltage.

The basic gain function is linear-in-dB, controlled by the voltage applied to Pin VDBS. The gain ranges from 0 dB to 50 dB for control voltages between 0 V and 1.5 V—a slope of 30 mV/dB. The gain linearity is typically within $\pm 0.1\text{ dB}$. By changing the logic level on Pin MODE, the gain decreases over the same range, with an opposite slope. A second gain control port is provided at the VMAG pin and allows the user to vary the numeric gain from a factor of 0.03 to 10. All the parameters of the AD8330 have low sensitivities to temperature and supply voltages.

¹ Protected by U.S. Patent No. 5,969,657; other patents pending.

Rev. D

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FUNCTIONAL BLOCK DIAGRAM

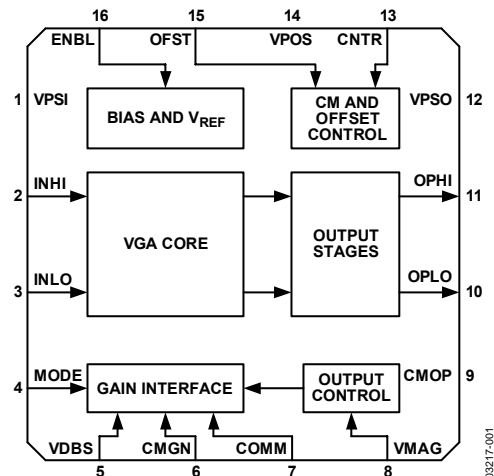


Figure 1.

Using VMAG, the basic 0 dB to 50 dB range can be repositioned to any value from 20 dB higher (that is, 20 dB to 70 dB) to at least 30 dB lower (that is, -30 dB to $+20\text{ dB}$) to suit the application, thereby providing an unprecedented gain range of over 100 dB. A unique aspect of the AD8330 is that its bandwidth and pulse response are essentially constant for all gains, over both the basic 50 dB linear-in-dB range, but also when using the linear-in-magnitude function. The exceptional stability of the HF response over the gain range is of particular value in those VGA applications where it is essential to maintain accurate gain law-conformance at high frequencies.

An external capacitor at Pin OFST sets the high-pass corner of an offset reduction loop, whose frequency can be as low as 5 Hz. When this pin is grounded, the signal path becomes dc-coupled. When used to drive an ADC, an external common-mode control voltage at Pin CNTR can be driven to within 0.5 V of either ground or V_S to accommodate a wide variety of requirements. By default, the two outputs are positioned at the midpoint of the supply, $V_S/2$. Other features, such as two levels of power-down (fully off and a hibernate mode), further extend the practical value of this exceptionally versatile VGA.

The AD8330 is available in 16-lead LFCSP and 16-lead QSOP packages and is specified for operation from -40°C to $+85^\circ\text{C}$.

SPECIFICATIONS

$V_S = 5\text{ V}$, $T_A = 25^\circ\text{C}$, $C_L = 12\text{ pF}$ on OPHI and OPLO, $R_L = \infty$, $V_{DBS} = 0.75\text{ V}$, $V_{MODE} = \text{high}$, $V_{MAG} = \text{Pin VMAG open circuit (0.5 V)}$, $V_{OFST} = 0\text{ V}$, differential operation, unless otherwise noted.

Table 1.

Parameter	Conditions	Min	Typ	Max	Unit
INPUT INTERFACE	Pin INHI, Pin INLO				
Full-Scale Input	$V_{DBS} = 0\text{ V}$, differential drive	± 1.4	± 2		V
Input Resistance	$V_{DBS} = 1.5\text{ V}$ Pin-to-pin	± 4.5	± 6.3		mV
Input Capacitance	Either pin to COMM	800	1 k	1.2 k	Ω
Voltage Noise Spectral Density	$f = 1\text{ MHz}$, $V_{DBS} = 1.5\text{ V}$; inputs ac-shorted		4		pF
Common-Mode Voltage Level			5		$\text{nV}/\sqrt{\text{Hz}}$
Input Offset	Pin OFST connected to Pin COMM		3.0		V
Drift			1		mV rms
Permissible CM Range ¹		0	2	V_S	$\mu\text{V}/^\circ\text{C}$
Common-Mode AC Rejection	$f = 1\text{ MHz}$, 0.1 V rms $f = 50\text{ MHz}$		-60 -55		V dB dB
OUTPUT INTERFACE	Pin OPHI, Pin OPLO				
Small Signal -3 dB Bandwidth	$0\text{ V} < V_{DBS} < 1.5\text{ V}$		150		MHz
Peak Slew Rate	$V_{DBS} = 0\text{ V}$		1500		$\text{V}/\mu\text{s}$
Peak-to-Peak Output Swing	$V_{MAG} \geq 2\text{ V}$ (peaks are supply limited)	± 1.8	± 2	± 2.2	V
Common-Mode Voltage	Pin CNTR O/C	± 4	± 4.5		V
Voltage Noise Spectral Density	$f = 1\text{ MHz}$, $V_{DBS} = 0\text{ V}$	2.4	2.5	2.6	V
Differential Output Impedance	Pin-to-pin	120	150	180	Ω
HD2 ²	$V_{OUT} = 1\text{ V p-p}$, $f = 10\text{ MHz}$, $R_L = 1\text{ k}\Omega$		-62		dBc
HD3 ²	$V_{OUT} = 1\text{ V p-p}$, $f = 10\text{ MHz}$, $R_L = 1\text{ k}\Omega$		-53		dBc
OUTPUT OFFSET CONTROL	Pin OFST				
AC-Coupled Offset	C_{HPF} on Pin OFST ($0\text{ V} < V_{DBS} < 1.5\text{ V}$)		10		mV rms
High-Pass Corner Frequency	$C_{HPF} = 3.3\text{ nF}$, from OFST to CNTR (scales as $1/C_{HPF}$)		100		kHz
COMMON-MODE CONTROL	Pin CNTR				
Usable Voltage Range		0.5		4.5	V
Input Resistance	From Pin CNTR to $V_S/2$		4		$\text{k}\Omega$
DECIBEL GAIN CONTROL	V_{DBS} , CMGN, and MODE pins				
Normal Voltage Range	CMGN connected to COMM		0 to 1.5		V
Elevated Range	CMGN O/C (V_{CMGN} rises to 0.2 V)		0.2 to 1.7		V
Gain Scaling	Mode high or low	27	30	33	mV/dB
Gain Linearity Error	$0.3\text{ V} \leq V_{DBS} \leq 1.2\text{ V}$	-0.35	± 0.1	+0.35	dB
Absolute Gain Error	$V_{DBS} = 0\text{ V}$	-2	± 0.5	+2	dB
Bias Current	Flows out of Pin V_{DBS}		100		nA
Incremental Resistance			100		$\text{M}\Omega$
Gain Settling Time to 0.5 dB Error	V_{DBS} stepped from 0.05 V to 1.45 V or 1.45 V to 0.05 V		250		ns
Mode Up/Down	Pin MODE				
Mode Up Logic Level	Gain increases with V_{DBS} , MODE = O/C	1.5			V
Mode Down Logic Level	Gain decreases with V_{DBS}			0.5	V
LINEAR GAIN INTERFACE	Pin VMAG, Pin CMGN				
Peak Output Scaling, Gain vs. V_{MAG}	See the Circuit Description section	3.8	4.0	4.2	V/V
Gain Multiplication Factor vs. V_{MAG}	Gain is nominal when $V_{MAG} = 0.5\text{ V}$		$\times 2$		
Usable Input Range		0		5	V
Default Voltage	V_{MAG} O/C	0.48	0.5	0.52	V
Incremental Resistance			4		$\text{k}\Omega$
Bandwidth	For $V_{MAG} \geq 0.1\text{ V}$		150		MHz

AD8330

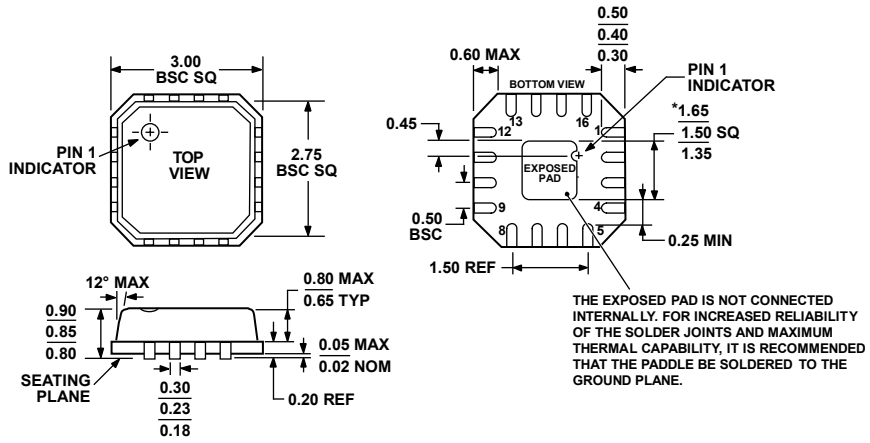
Parameter	Conditions	Min	Typ	Max	Unit
CHIP ENABLE	Pin ENBL			0.5	V
Logic Voltage for Full Shutdown					
Logic Voltage for Hibernate Mode	Output pins remain at CNTR	1.3	1.5	1.7	V
Logic Voltage for Full Operation		2.3			V
Current in Full Shutdown			20	100	μ A
Current in Hibernate Mode			1.5		mA
Minimum Time Delay ³			1.7		μ s
POWER SUPPLY	VPSI, VPOS, VPSO, COMM, and CMOP pins				
Supply Voltage		2.7		6	V
Quiescent Current	$V_{DBS} = 0.75$ V		20	27	mA

¹ The use of an input common-mode voltage significantly different from the internally set value is not recommended due to its effect on noise performance. See Figure 56.

² See the Typical Performance Characteristics section for more detailed information on distortion in a variety of operating conditions.

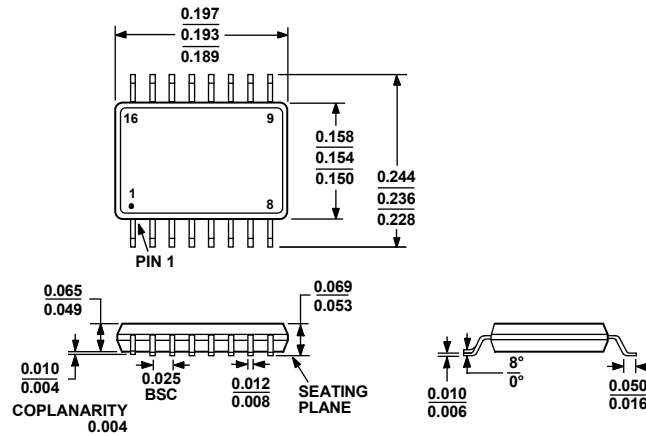
³ For minimum sized coupling capacitors.

OUTLINE DIMENSIONS



*COMPLIANT TO JEDEC STANDARDS MO-220-VEED-2 EXCEPT FOR EXPOSED PAD DIMENSION.

Figure 78. 16-Lead Lead Frame Chip Scale Package [LFCSP_VQ]
3 mm x 3 mm Body, Very Thin Quad
(CP-16-3)
Dimensions shown in millimeters



COMPLIANT TO JEDEC STANDARDS MO-137-AB
Figure 79. 16-Lead Shrink Small Outline Package [QSOP]
(RQ-16)
Dimensions shown in inches

ORDERING GUIDE

Model	Temperature Range	Package Description	Package Option	Branding
AD8330ACP-R2	−40°C to +85°C	16-Lead LFCSP_VQ	CP-16-3	JFA
AD8330ACP-REEL	−40°C to +85°C	16-Lead LFCSP_VQ	CP-16-3	JFA
AD8330ACP-REEL7	−40°C to +85°C	16-Lead LFCSP_VQ	CP-16-3	JFA
AD8330ACPZ-R2 ¹	−40°C to +85°C	16-Lead LFCSP_VQ	CP-16-3	JFZ
AD8330ACPZ-RL ¹	−40°C to +85°C	16-Lead LFCSP_VQ	CP-16-3	JFZ
AD8330ACPZ-R7 ¹	−40°C to +85°C	16-Lead LFCSP_VQ	CP-16-3	JFZ
AD8330ARQ	−40°C to +85°C	16-Lead QSOP	RQ-16	
AD8330ARQ-REEL	−40°C to +85°C	16-Lead QSOP	RQ-16	
AD8330ARQ-REEL7	−40°C to +85°C	16-Lead QSOP	RQ-16	
AD8330ARQZ ¹	−40°C to +85°C	16-Lead QSOP	RQ-16	
AD8330ARQZ-RL ¹	−40°C to +85°C	16-Lead QSOP	RQ-16	
AD8330ARQZ-R7 ¹	−40°C to +85°C	16-Lead QSOP	RQ-16	
AD8330-EVAL		Evaluation Board		
AD8330-EVALZ ¹		Evaluation Board		

¹ Z = RoHS Compliant Part.